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SC-FDMA-based resource allocation and power control scheme for D2D communication using LTE-A uplink resource

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Abstract

Device-to-device (D2D) communication-enabled cellular networks allow cellular devices to directly communicate with each other without any evolved NodeB (eNB). D2D communication aims to improve the spectral efficiency and increases the overall system capacity. For future mobile networks, intelligent radio resource allocation and power control schemes are required to accommodate the increasing number of cellular devices and their growing demand of data traffic. In this paper, a combined resource allocation and power control scheme for D2D communication is proposed. In the proposed scheme, D2D communication reuses the uplink (UL) resources of conventional cellular user equipments (CUEs); therefore, we have adopted single-carrier frequency division multiple access (SC-FDMA) as UL transmission scheme. The proposed scheme uses fractional frequency reuse (FFR)-based architecture to efficiently allocate the resources and mitigate the interference between CUEs and D2D user equipments (DUEs). In order to guarantee the user fairness, the proposed scheme uses the well-known proportional fair (PF) scheduling algorithm for resource allocation. We have also proposed an intelligent power control scheme which provides equal opportunity to both CUEs and DUEs to achieve a certain minimum signal-to-interference and noise ratio (SINR) value. The performance evaluation results show that the proposed scheme significantly improves the overall cell capacity and achieves low peak-to-average power ratio (PAPR).

Keywords: D2D communication; SC-FDMA; Resource allocation; Power control; PF scheduling

1 Introduction

Recent years have witnessed rapid increase in the demand of ubiquitous communication services [1,2]. Because of their almost ubiquitous nature, cellular networks have been massively deployed and used ever since their introduction in 1980s. Several cellular technologies have surfaced over the past some time. Most recently, the long-term evolution (LTE) and its successor LTE-advanced (LTE-A) have attracted considerable attention in the cellular domain. LTE-A is essentially an improvement on LTE, which aims at offering a variety of new techniques such as carrier aggregation, multiple-input multiple-output (MIMO), device-to-device (D2D) communication, etc.

Unlike conventional cellular technology that conveys information between two user equipments (UEs) via evolved NodeB (eNB), D2D communications allows UE-to-UE data exchange on direct links. By relieving the eNBs from too much relaying, D2D communications improves the overall system capacity and end user data rates [3]. The idea is that the eNB shall help the two UEs in establishing a direct connection between them, and then let the communication session evolve on one-hop D2D link. The UEs that are capable of communicating directly are referred to as D2D UEs (or DUEs). In certain situations, it may be imperative for the UEs to send information through the eNB. For instance, if the distance between two UEs is very large, direct communication cannot take place. Therefore, a cellular network that supports D2D communications has two kinds of UEs. One is the DUEs that transmit and receive data without eNB relaying, and the other is conventional cellular UEs (or CUEs) that communicate via the base station [4]. The 3rd Generation Partnership

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Project (3GPP) has considered two modes of D2D communications: mode-1 and mode-2 [5]. In mode-1 D2D communications, D2D users are located in an eNB coverage zone (i.e., in coverage). Thus, the eNB is responsible for D2D connection establishment, resource allocation, monitoring, and so on. On the other hand, in mode-2 D2D communications, D2D users are not located in the coverage zone of an eNB. Therefore, D2D users select radio resources autonomously. For D2D service scenarios such as content sharing, proximity-based advertisements, etc., Mode-1 communications is preferred [6-9]. In this paper, we focus on mode-1 D2D communications. Figure 1 shows an example of D2D communications underlying cellular network.

It is intuitive that D2D communications under 3GPP mode-1 category shall result in severe interference if CUEs and DUEs are operating on the same frequency band. This so-called underlay technique requires a systematic frequency reuse strategy in order to prove beneficial [3,10]. The other technique that allows CUEs and DUEs to coexist within a cell is the overlay technique. The overlay technique splits the frequency band allocated to a cell and distributes it among the CUEs and DUEs. In the overlay mode, CUEs and DUEs operate at two different frequency sub-bands. While the overlay technique works best in reducing interference, it results in the underutilization of the frequency resource [3,10]. Almost all previous works on D2D communications prioritize the underlay technique because it can accommodate more UEs on a given frequency band. In order to ensure that interference

is avoided in the underlay technique, various previous works have focused on proposing techniques that (i) intelligently allocate frequency resources to CUEs and DUE [11] and (ii) smartly control the transmit power of CUEs and DUEs [12]. This paper focuses on proposing a combined method that allows CUEs and DUEs to exist on the same frequency band in a typical LTE-A network.

LTE-A networks use single-carrier frequency division multiple access (SC-FDMA) for uplink transmissions [13]. An SC-FDMA system uses different subchannels to transmit the information symbols in parallel with each other. Compared with OFDMA, SC-FDMA significantly reduces the envelope fluctuations in the transmitted waveform. Thus, SC-FDMA signals have inherently lower peak-to-average power ratio (PAPR) than the OFDMA signals. In SC-FDMA, throughput depends on how the information symbols are applied to the subchannels. There are two possible implementation approaches to SC-FDMA: SC-FDMA with constraints (also known as localized SC-FDMA) and distributed SC-FDMA. In SC-FDMA with constraints, each terminal uses a set of adjacent subcarriers to transmit its symbols. On the other hand, in distributed SC-FDMA, the subcarriers are spread over the entire signal band. It has been shown in [14-16] that SC-FDMA with constraints achieves higher throughput than the distributed SC-FDMA. Therefore, this work employs SC-FDMA with constraints for uplink transmissions. For this kind of SC-FDMA, it has two critical constraints in relation to resource allocation. The first is exclusivity which maintains that a subchannel can only be allocated

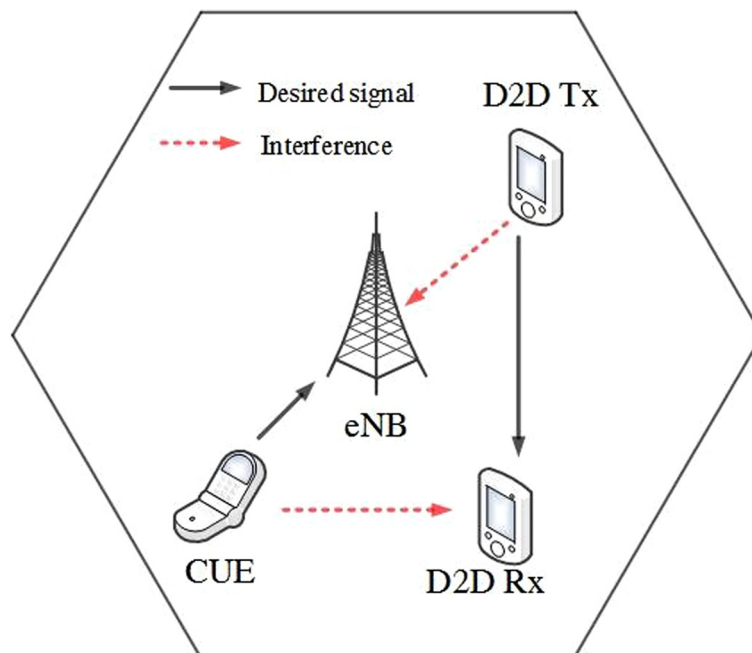


Figure 1 D2D communications. DUEs reusing the UL resource of CUEs in D2D communications underlying cellular network.

to one UE. The second constraint is adjacency, which states that if multiple subchannels are assigned to a UE, all of them must be adjacent to each other in the frequency domain.

1.1 Contribution

Various studies [16-20] have addressed the exclusivity and adjacency constraints for proportionally fair (PF) scheduling-based resource allocation. However, all previous works consider only CUEs and do not consider D2D communications. Hence, PF schedulers used in [16-20] are only for single-tier network where only CUEs exist. In order to apply the resource allocation to both CUEs and DUEs using uplink resources, our previous work has considered SC-FDMA-based resource allocation scheme for D2D communications [21]. Although this scheme avoids the interference between CUEs and DUEs by using the fractional frequency reuse (FFR) concept, it does not guarantee minimum performance requirements for both CUEs and DUEs. For satisfying the minimum SINR requirements, this work enhances the previous scheme with a novel power control technique in conjunction with PF scheduling and SC-FDMA-based resource allocation. To the best of our knowledge, this is the first work that proposes combined power control and resource allocation scheme based on constrained SC-FDMA and PF scheduling. In this paper, we show that our proposed scheme significantly improves the overall network performance.

The rest of the paper is organized as follows. Section 2 covers literature review which highlights the works done in the context of frequency allocation and transmit power control. The previous works that address combined resource allocation and power control have also been discussed. Section 3 explains the FFR concept and the proposed radio resource allocation and transmit power control scheme for CUEs and DUEs. The observations on performance evaluation have been reported in Section 4, and this paper is concluded in Section 5.

2 Related work

Considerable research has been conducted on spectrum sharing between CUEs and DUEs and their transmit power control. Some of the relevant works are briefly highlighted in the following.

2.1 Transmit power control

A power control and link selection algorithm with temporary removal of DUEs is proposed in [22]. A single-cell CDMA cellular system is considered, where the transmit power of the DUEs is minimized through link selection. The temporary removal algorithm sets the transmit power of DUEs to zero when the required power exceeds the predefined maximum power. Gabor et al. have studied the performance of various power control strategies in

LTE networks that are applicable to D2D communications [23]. The performance of LTE power control scheme and a utility function-based distributed power control scheme are compared, where dynamic resource allocation and mode selection are implemented. In [24], Yu et al. have analyzed different power control schemes for D2D communications in a cellular network with orthogonal and non-orthogonal resource sharing. The study focuses on network sum rate maximization by using power control mechanism. Moreover, in a different scenario, CUEs have been prioritized and an upper bound is placed on the maximum transmission rate of all links. It is concluded that reasonable quality of service can be guaranteed with increased sum rate by using rate-constrained power control mechanism. Gu et al. in [25] have proposed a dynamic power control scheme for D2D communications. The scheme excludes eNB, CUEs using the same resource, and the areas of adjacent cells from the coverage of D2D communications. This is achieved by periodically adjusting the transmit power of the DUEs. It has been shown that the proposed scheme outperforms the conventional power control schemes. In [26], four different kinds of power control schemes for DUEs are proposed. The authors conclude that the closed loop power control with a dynamic tuning step may be suitable for D2D communications. It has also been concluded that power control alone is not an effective method to avoid co-channel interference between CUEs and DUEs.

Most previous works on power control prioritize CUEs by guaranteeing a certain signal-to-interference and noise ratio (SINR). However, our proposed transmit power control scheme also provides equal opportunity to DUEs to maintain a reasonable SINR at all times.

2.2 Resource allocation

Hyang et al. in [27] have proposed a frequency resource allocation scheme for D2D communications in cellular networks using strict FFR. It has been shown that an improved SINR can be achieved by reusing the downlink resources such that a certain spatial distance is maintained between CUEs and the interfering DUEs. Lee et al. have proposed a PF scheduling-based resource allocation scheme that improves the throughput fairness among all users [28]. The scheme forms a pair between a CUE and a D2D pair for sharing the same resource. This resource sharing pair is formed on the basis of PF metric value of CUEs and the interference level between the pair. Doppler et al. in [29] propose mode selection algorithm for D2D communications underlying cellular network. In this work, eNB decides whether (i) D2D pairs get dedicated resource, (ii) reuse the cellular resource, or (iii) communicate via eNB. The paper concludes that by using the proposed mode selection, the network sum rate is improved by 50% and the ratio of successful D2D communications

is more than doubled. In [30], authors have proposed a distance-constrained resource sharing scheme for underlaying D2D communications, which identifies an optimal minimum distance between resource sharing D2D pair and the CUE. The scheme mitigates the interference between CUEs and DUEs and reduces the outage probability of D2D communications. Keeping in mind the adjacency and exclusivity constraint of SC-FDMA [31], Zhang et al. propose a two-step greedy resource allocation algorithm which assigns contiguous resources to CUEs [20]. This has been done by maintaining the long-term user fairness and spectral efficiency of the network. In [32], the authors have studied PF frequency domain packet scheduling (FDPS) with contagious resource allocation constraint in SC-FDMA. A set of algorithms have been proposed to solve the adjacency constraint in SC-FDMA and to find the near-to-optimal scheme which best emulates the time domain PF criteria. Wong et al. in [17] have proposed an optimal algorithm to assign contiguous resources to CUEs in SC-FDMA-based system. The study mainly addresses the adjacency constraint in SC-FDMA. The proposed algorithm is a reformulation of the problem as a pure binary integer program called the set partitioning problem. It concludes that the proposed algorithm can maximize the logarithmic sum rate of a SC-FDMA-based system.

It has been argued before that most of the previous works have either used OFDMA in resource allocation for CUEs and DUEs or have used SC-FDMA for CUEs only. Thus, in our preliminary version of this work [21], we have proposed a SC-FDMA-based resource allocation scheme for D2D communications. However, the work only considers the resource allocation, and it does not guarantee the SINR of CUEs and DUEs according to their minimum SINR requirements. In order to guarantee the minimum SINR requirements, in this extended work of our previous study, we propose a combined resource allocation and power control mechanism for D2D communications.

2.3 Combined power control and resource allocation

Janis et al. propose a resource allocation scheme [33] and a power control scheme [10] for CUEs and DUEs working in an underlay cellular network. The scheme reallocates the CUE resources to DUEs in such a manner that it causes least interference to CUEs. Moreover, resources are allocated to D2D pair if it satisfies a certain minimum SINR threshold for CUEs. In [34], the authors have investigated a joint power control and radio resource allocation scheme for D2D communications underlaying cellular networks. An optimization problem for energy efficiency has also been formulated. The energy efficiency of the system has been maximized using an iterative combinatorial auction game approach. Gu et al. in [35] have proposed a combined power control and resource allocation scheme

for D2D communications in an OFDMA-based cellular system. The power control scheme ensures a certain threshold SINR for CUEs and imposes a maximum limit on SINR of DUEs. Upon comparison with [33], it has been shown that a better sum rate can be achieved by using the proposed scheme.

In [36], the authors propose an iterative algorithm which jointly optimizes the matching and power control of CUEs and D2D links. The work in [36] also investigates the reuse of downlink radio resource between multiple D2D links and multiple CUEs. Moreover, the scheme imposes a QoS target for each CUE link to ensure better performance of the cellular network. It has been shown that by proper matching and power coordination, an optimal resource reuse solution can be achieved. It has been highlighted in [37] that reusing UL and DL resources for D2D communications has different pros and cons in different environments. The conclusion is that reusing DL can improve DUE SINR if they are located closer to the eNB. On the other hand, UL is effective for reuse when DUEs are farther away from eNB.

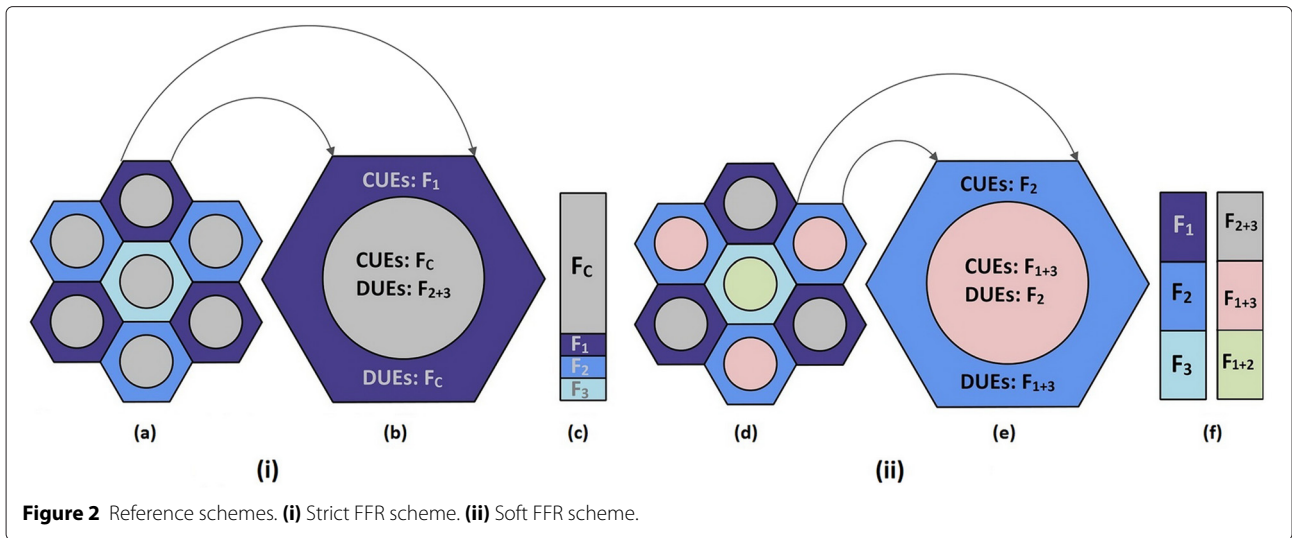
3 Proposed scheme

This section starts with an introduction to the proposed network architecture. A brief review of FFR has been given first. The resource allocation and transmit power control schemes are also proposed later in this section.

3.1 Network architecture

In a typical cellular network, it is possible to divide a cell into inner (or center) and outer zones such that different sub-bands are allocated to each zone. According to strict FFR [38,39], the center zone has a frequency reuse factor (FRF) of 1 while the outer zones have FRF of N . Figure 2a,b,c shows the deployment of a network using strict FFR with $N = 3$ at the outer zone. It can be seen from the figures that DUEs in the center zone reuse the frequency of the outer zone of the neighboring cells. It can also be observed from Figure 2a,b,c that strict FFR does not consume the entire frequency band in a single cell. On the other hand, a network employing soft FFR uses the entire spectrum in a single cell and reuses the same spectrum for the neighboring cells. Figure 2d,e shows a network deployment that uses soft FFR with $N = 3$ in the outer zone. The entire frequency spectrum is divided into three partitions such that one partition is a combination of two separate sub-bands. The CUEs in the center zone use the frequency resources that are not used in the outer region of the same cell (F1 and F3 in Figure 2e). On the other hand, DUEs can use the frequency resources that are not currently being used by the CUEs residing in the same region (F2 in Figure 2e).

In the proposed network architecture, each cell is divided into four different sectors S_C , S_1 , S_2 , and S_3 as



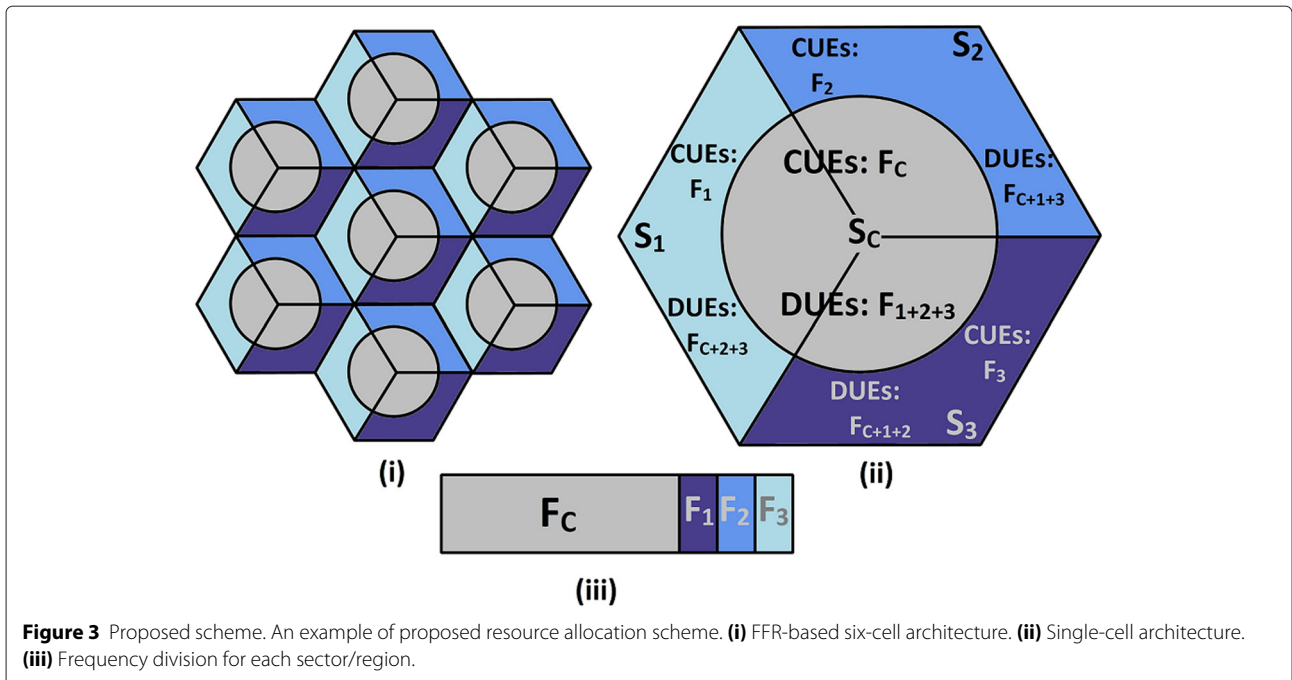
shown in Figure 3. The frequency band allocated to a cell is also divided into four partitions $F_C, F_1, F_2,$ and F_3 such that each frequency partition is allocated to each of the four sectors. Note that the central sector, S_C , covers the largest area within the cell and hence gets a larger partition F_C . Thus, one of the important design parameters is the radius of S_C within a cell. In this work, we assume S_C to be 63% of the total cell area [40].

3.2 System model

The resource allocation scheme proposed here uses FFR for D2D communications underlying cellular network operating in TDD mode. The proposed scheme uses TTD

configuration 1 because it is the only configuration that can support symmetric downlink and uplink transmissions [41]. The uplink resources of CUEs are shared by the DUEs under the constraints of SC-FDMA highlighted earlier in Section 1. The main advantage of reusing the uplink resources is that eNB can effectively coordinate interference between CUEs and DUEs in comparison with the case where downlink resources are reused [42,43]. The proposed scheme works as follows.

Consider a cell with N_C CUEs and N_D D2D pairs. We define $\mathbb{C} = \{c|c = 1, 2, \dots, N_C\}$ and $\mathbb{D} = \{d|d = 1, 2, \dots, N_D\}$ as the sets of CUEs and DUEs in the cell. Each D2D pair consists of a D2D transmitter Tx and its

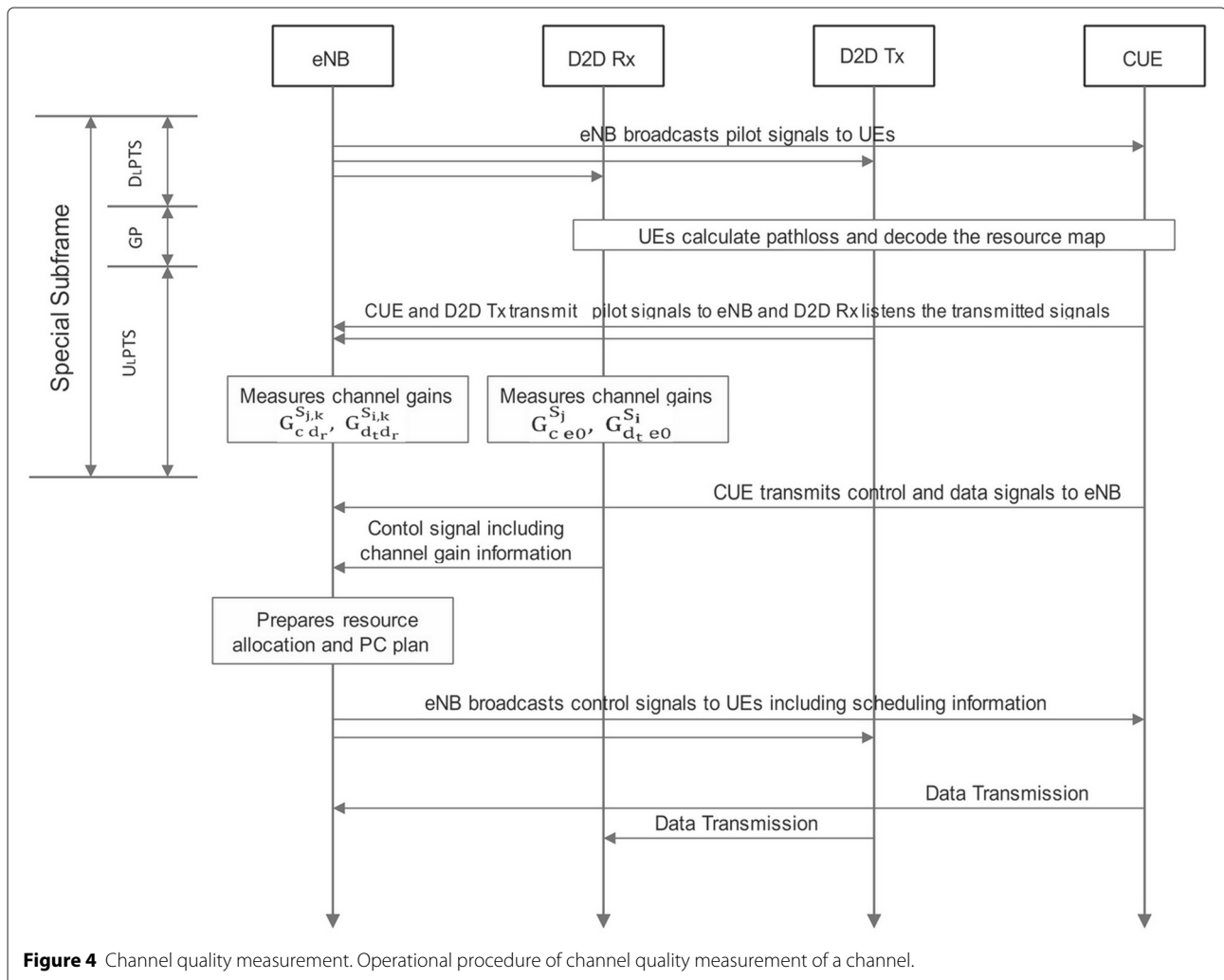


corresponding D2D receiver Rx. A CUE residing in any of the four sectors can only use the frequency band allocated to that sector only. For example, a CUE in sector S_C shall use the resource partition F_C . However, a DUE can use any frequency resource except the one allocated to its sector. We assume that the information on which sector a certain UE is located can be determined by using the location services (LCS) as explained in 3GPP Rel-9 [44]. We also define a resource sharing pair f_{cd} ($c \in N_C$ and $d \in N_D$) that consists of one CUE and one D2D pair. Note that the CUE and the D2D pair constituting f_{cd} are located in different sectors. A typical example of the proposed resource allocation has been shown in Figure 3. It has been shown that a D2D pair in sector S_1 can only use one of the frequency sub-bands $F_C, F_2,$ and F_3 but not F_1 . Thus, each DUE can choose among three frequency sub-bands that are already being used by their respective CUEs.

In order to allocate appropriate frequency resources to the UEs, eNB measures the channel quality between itself and the UEs. Figure 4 shows the operational procedure

of channel quality measurement. The procedure starts when eNB gathers information on channel gains of all UEs (CUEs and D2D receivers) in a cell. In the downlink pilot time slot of the special sub-frame, eNB sends pilot signals to all UEs. These pilot signals contain the resource map of the uplink pilot channels allocated to the UEs. All UEs calculate the path loss between themselves and eNB based on the received strength of the downlink pilot signal. UEs transmit a pilot signal back to eNB using the uplink pilot time slot. eNB uses the strength of this signal to measure the channel gain between itself and the UE.

The D2D receivers (Rx) also measure the channel gain between themselves and the CUEs, as well as the gain between themselves and the D2D transmitters (Tx). D2D Rx listens to the uplink pilot signals sent by the CUEs and D2D Tx in order to determine the channel gains. Figure 4 shows the notation for different channel gains between CUEs, D2Ds, and eNB. G_{tr}^{ij} indicates the gain between transmitter t located in sector i and the receiver r located



in sector j . As can be seen from Figure 5, $G_{ce0}^{S_j}$ represents the gain between a CUE c and eNB $e0$ and $G_{d_t d_r}^{S_i, k}$ is the gain between DUE Tx and its DUE Rx such that Tx is in sector i and Rx is in sector k , and so on. It is obvious that i, j , and $k \in S_C, S_1, S_2$, and S_3 .

3.3 Problem formulation

For allocating resources to CUEs and D2D pairs under the constraints of SC-FDMA, eNB first allocates contiguous subchannels to CUEs [16,20]. To begin with, eNB divides the whole bandwidth into several sub-channels. eNB then chooses one CUE for each sub-channel based on the highest PF metric value. The main target of PF scheduling is to maximize the logarithmic utility function $\sum_u \log R_u$, where R_u is the historical average data rate of user u . The maximization of $\sum_u \log R_u$ can be transformed equivalently to the maximization of $\sum_u \log \frac{r_u}{R_u}$ at each transmit time interval (TTI), where r_u represents the achievable data rate of user u at the current TTI. We define $\lambda_{u,n} = \frac{r_{u,n}}{R_u}$ as the PF metric value for user u on subchannel n . The PF metric in this work is calculated on a per-sector basis as follows:

$$\lambda_{c,n}^{S_j} = \frac{r_{c,n}^{S_j}}{R_c^{S_j}}, \tag{1}$$

where $r_{c,n}^{S_j}$ denotes the achievable data rate of CUE (located in sector j) in current TTI on subchannel n (where $n \in F_C, F_1, F_2$, or F_3) and $R_c^{S_j}$ represents the average data rate of CUE in the previous transmission intervals. The optimization problem for the proposed SC-FDMA-based resource allocation scheme with adjacency and exclusivity constraints using multi-tier PF scheduler [45-48] can be formulated as:

$$\max_{\substack{\{A_1, \dots, A_{N_C}\} \in \mathbb{K} \\ \text{and } \{\mathbb{B}_1, \dots, \mathbb{B}_{N_D}\} \in \mathbb{K}}} \left(\sum_{c \in \mathbb{C}} \log \left(\sum_{n \in A_c} \lambda_{c,n} \right) + \sum_{d \in \mathbb{D}} \log \left(\sum_{n \in \mathbb{B}_d} \lambda_{d,n} \right) \right) \tag{2}$$

subject to $A_c \cap A_{c'} = \emptyset, \forall c \neq c', \text{ for } c, c' \in \mathbb{C}$
 $\mathbb{B}_d \cap \mathbb{B}_{d'} = \emptyset, \forall d \neq d', \text{ for } d, d' \in \mathbb{D}$
 $P_{c,n} = P_c / |A_c|, \text{ for } n \in A_c$
 $P_{d_t, n} = P_{d_t} / |\mathbb{B}_d|, \text{ for } n \in \mathbb{B}_d$.

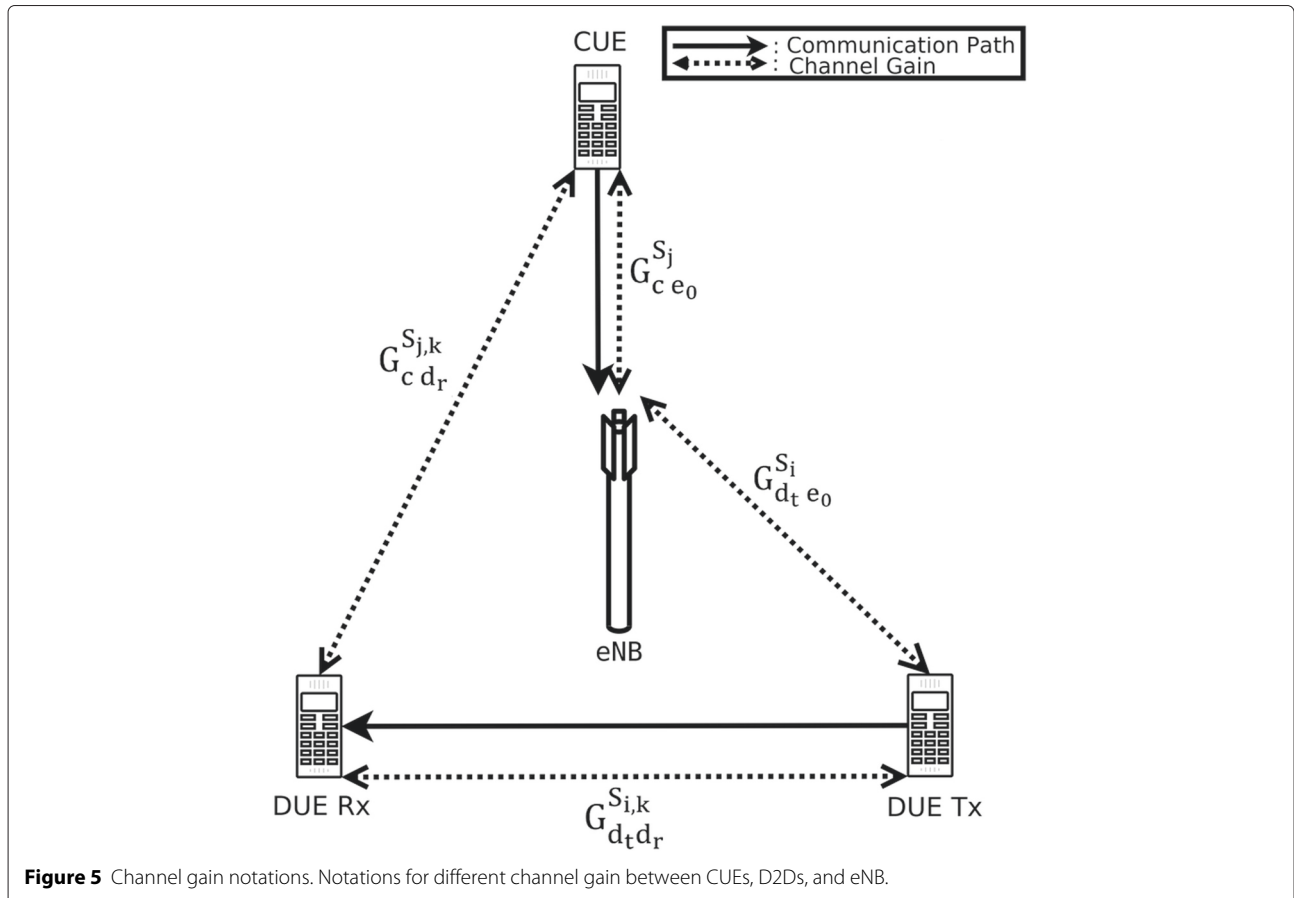


Figure 5 Channel gain notations. Notations for different channel gain between CUEs, D2Ds, and eNB.

For simplicity, Equation 2 can also be written in the product form, as follows:

$$\max_{\substack{\{\mathbb{A}_1, \dots, \mathbb{A}_{N_C}\} \in \mathbb{K} \\ \text{and } \{\mathbb{B}_1, \dots, \mathbb{B}_{N_D}\} \in \mathbb{K}}} \left(\prod_{c \in \mathbb{C}} \log \left(\sum_{n \in \mathbb{A}_c} \lambda_{c,n} \right) \cdot \prod_{d \in \mathbb{D}} \log \left(\sum_{n \in \mathbb{B}_d} \lambda_{d,n} \right) \right), \quad (3)$$

where \mathbb{K} is the set of all possible subchannel allocations satisfying the adjacency constraint. \mathbb{A}_c (and \mathbb{B}_d) and $|\mathbb{A}_c|$ (and $|\mathbb{B}_d|$), respectively, represent the set of allocated subchannels and the number of allocated subchannels to CUE c (and DUE d). The first and the second constraints show the exclusivity constraints for CUEs and DUEs, respectively. In other words, a subchannel cannot be allocated to any other CUE and D2D pair (f_{cd}) in the same TTI, if it is already allocated to a CUE and a D2D pair. Similarly, the third and the fourth constraints mean that the CUEs and DUEs are using equal transmit power on all allocated subchannels, where P_c and P_{d_t} are the transmit power of CUE c and DUE Tx d_t , respectively. The notations used in this paper are summarized in Table 1. The formulated objective problem (2) is NP hard in nature, since it is applied on all subchannels and the set \mathbb{A}_c do not have the same length.

3.4 Proposed resource allocation

Finally, eNB allocates the resources to DUEs under the constraints of SC-FDMA such that the interference between D2D Tx and eNB is mitigated. For resource allocation to DUEs, eNB calculates the PF metric values of CUEs and each D2D pair as follows:

$$\lambda_{c,n}^{S_j} = 1 + \frac{r_{c,n}^{S_j}}{R_c^{S_j}} \quad (4)$$

and

Table 1 Definition of notations

Notations	Definition
N_C	The number of CUEs per cell
N_D	The number of DUEs per cell
\mathbb{C}	The set of CUEs in the cell
\mathbb{D}	The set of DUEs in the cell
$\lambda_{c,n}$	PF metric value of CUE c on subchannel n
$\lambda_{d,n}$	PF metric value of DUE d on subchannel n
$G_{d_t e0}$	Channel gain between D2D Tx d_t and eNB $e0$
$G_{d_t d_r}$	Channel gain between D2D Tx d_t and D2D Rx d_r
G_{ce0}	Channel gain between CUE c and eNB $e0$
G_{cd_r}	Channel gain between CUE c and D2D Rx d_r
N_0	Noise power

$$\lambda_{d,n}^{S_{i,k}} = 1 + \frac{r_{d,n}^{S_{i,k}}}{P_d^{S_{i,k}}}, \quad (5)$$

where $r_{c,n}^{S_j}$ and $r_{d,n}^{S_{i,k}}$ denotes the achievable data rate of a CUE c and a DUE d (located in different sectors) on subchannel n . In order to consider the interference between resource sharing pairs, eNB calculates the product of both PF metric values ($\lambda_{c,n}^{S_j} \times \lambda_{d,n}^{S_{i,k}}$). Similarly, eNB shall form the resource sharing pair f_{cd} consisting of a CUE and D2D pair if the value of ($\lambda_{c,n}^{S_j} \times \lambda_{d,n}^{S_{i,k}}$) is the highest for this pair among all other combinations of CUEs and DUEs. The resource sharing pairs comprising of the rest of the CUEs and DUEs are also formed in the similar manner. Moreover, the proposed resource allocation scheme has computational complexity of $O(N_C \times N_K^2)$ for computing PF metric values for all CUEs on all subchannels. Likewise, computational complexity for allocating resources to DUEs is $O(N_D \times N_K)$, where N_C and N_D are the number of CUEs and DUEs, respectively. N_K is the number of subchannels. Since resource allocation is performed in parallel, the computational complexity of the entire proposed algorithm is $O((N_C \times N_K^2) + (N_D \times N_K))$. This increase is natural because eNB has to perform additional computations to allocate resources to DUEs.

3.5 Determining transmit power of DUEs

In the proposed scheme, CUEs use fixed transmit power while the transmit power of DUEs is controlled based on the network condition. We assume that enough subchannels are available in each sector to support all CUEs. The proposed power control scheme for DUEs is such that it provides equal opportunity to both CUEs and DUEs to maintain a certain SINR level. eNB calculates the SINR of the resource sharing pair f_{cd} using:

$$\text{SINR}_C^{S_j} = \frac{P_c^{S_j} G_{ce0}^{S_j}}{N_0 + P_{d_t}^{S_i} G_{d_t e0}^{S_i}} \quad (6)$$

and

$$\text{SINR}_{D2D}^{S_{i,k}} = \frac{P_{d_t}^{S_i} G_{d_t d_r}^{S_{i,k}}}{N_0 + P_c^{S_j} G_{cd_r}^{S_{j,k}}}, \quad (7)$$

where N_0 is the thermal noise power and $P_c^{S_j}$ is the transmit power of CUE c in sector j . $P_{d_t}^{S_i}$ is the transmit power of DUE Tx located in sector i . According to the proposed resource allocation scheme, $i \neq j$ and $i = k$ if both D2D Tx and Rx reside in the same sector. Since eNB already

knows the values of $N_0, P_c^{S_j}$ and the channel gains, it determines the transmit power $P_{d_t}^{S_i}$ of D2D Tx with in $[0, P_{\max}]$, where P_{\max} is the maximum possible transmit power of a UE.

Unlike most of the previous studies on power control [22-26] which only guarantee the SINR of CUEs, the proposed power control scheme not only guarantees a threshold SINR of CUEs γ_c but also provides equal opportunity to DUEs to maintain a certain a SINR threshold γ_d . If SINR of a CUE is expected to be lower than γ_c and D2D SINR is greater than or equal to γ_d , then DUE limits the interference by adjusting its transmit power. Similarly, if SINR of CUE is greater than γ_c and D2D SINR is less than γ_d , then DUE increases its transmit power such that SINR of CUE remains greater than γ_c . On other hand, if SINR of both CUE and DUE is greater than their respective threshold SINRs (γ_c and γ_d), DUE adjusts its transmit power to P_{\max} while making sure that SINR of CUE remains higher than γ_c .

In all cases where SINR of both CUE and DUE are expected to be less than their respective threshold values (γ_c and γ_d), the eNB decides whether to limit DUEs transmit power or to increase it. This decision is based on the difference between the SINR and threshold SINR values. For example, if the difference between CUE SINR and γ_c is less than that between DUE SINR and γ_d , DUE decreases its transmit power. On the other hand, if the difference between CUE SINR and γ_c is greater than that between DUE SINR and γ_d , DUE increases its transmit power. The idea here is that in the worst conditions, at least one communication session should proceed.

In summary, the transmit power of D2D Tx in f_{cd} is given by:

$$P_{d_t}^{S_i} = \begin{cases} \frac{P_c^{S_j} G_{ce0}^{S_j} - N_0}{G_{d_t e0}^{S_j} \gamma_c - G_{d_t e0}^{S_j}} & \text{if } \text{SINR}_C^{S_j} < \gamma_c \text{ and } \text{SINR}_{D2D}^{S_i,k} \geq \gamma_d \text{ or if } \min[\check{C}, \check{D}] = \check{C}, \\ \frac{\gamma_d P_c^{S_j} G_{cd_r}^{S_j,k} + \gamma_d N_0}{G_{d_t d_r}^{S_i,k} + G_{d_t d_r}^{S_i,k}} & \text{if } \text{SINR}_C^{S_j} \geq \gamma_c \text{ and } \text{SINR}_{D2D}^{S_i,k} < \gamma_d \text{ or if } \min[\check{C}, \check{D}] = \check{D}, \\ [P_{\max}] & \text{if } \text{SINR}_C^{S_j} \geq \gamma_c \text{ and } \text{SINR}_{D2D}^{S_i,k} \geq \gamma_d, \end{cases} \quad (8)$$

where \check{C} is the difference between CUE SINR and γ_c , while \check{D} is the difference between DUE SINR and γ_d , and can be defined as:

$$\check{C} = \text{SINR}_C^{S_j} - \gamma_c \quad (9)$$

and

$$\check{D} = \text{SINR}_{D2D}^{S_i,k} - \gamma_d. \quad (10)$$

To further analyze the proposed power control scheme in terms of D2D energy efficiency (D2D-EE), we have formulated and derived the optimal power control for DUEs as [49], Problem (P1) given in Equation 11. We compare

this optimal scheme with our proposed power control method in order to assess its performance. The aim of the formulated problem is to maximize the D2D energy efficiency while meeting the throughput requirements of both CUEs and DUEs. The optimization problem can be expressed as:

$$(P1) : \max_{P_c, P_{d_t}} \frac{1}{P_{d_t}} (R_c + R_d) \quad (11)$$

subject to

$$R_c = \log_2(1 + \text{SINR}_C) \geq R_c^{\text{req}} \quad (12)$$

$$R_d = \log_2(1 + \text{SINR}_{D2D}) \geq R_d^{\text{req}} \quad (13)$$

$$P_{\max C} \geq P_c > 0 \text{ and } P_{\max D} \geq P_{d_t} > 0. \quad (14)$$

In Problem (P1), Equation 11 represents the energy efficiency, where R_c, R_d , and P_{d_t} denotes throughput of CUEs, DUEs, and DUEs transmit power, respectively. Constraints given in Equations 12 and 13 enforce that the throughputs of both CUEs and DUEs should be respectively greater than or equal to the minimum throughput threshold R_c^{req} and R_d^{req} . Whereas constraint in Equation 14 ensures that the transmit power of both CUEs and DUEs should not exceed the maximum transmit power $P_{\max C}$ and $P_{\max D}$. In Problem (P1), the objective function given in Equation 11 is decreasing in both CUEs transmit power P_c and DUEs transmit power P_{d_t} . Therefore, in order to maximize Equation 11, both CUEs and DUEs should transmit with the lowest power (P_c and P_{d_t}) that ensure their threshold throughput requirements. Let P_c^* and $P_{d_t}^*$ indicate the optimal solution for Problem (P1). Note that the constraint given in Equations 12 and 13 are equivalent to:

$$\frac{P_c G_{ce0}}{N_0 + P_{d_t} G_{d_t e0}} \geq \gamma_c \text{ and } \frac{P_{d_t} G_{d_t d_r}}{N_0 + P_c G_{cd_r}} \geq \gamma_d, \quad (15)$$

respectively. Where γ_c and γ_d are the minimum threshold SINR for CUEs and DUEs, respectively, and can be expressed as $\gamma_c = 2^{R_c^{\text{req}}} - 1$ and $\gamma_d = 2^{R_d^{\text{req}}} - 1$. It is possible to obtain the minimum transmit power of both CUEs and DUEs by strictly binding and solving the constraints in Equation 15. Thus, the minimum transmit power (lower bound) of both CUEs and DUEs denoted as \check{P}_c and \check{P}_{d_t} can be expressed as:

$$\check{P}_c = \frac{\gamma_c N_0 (\gamma_d G_{d_t e0} + G_{d_t d_r})}{G_{d_t d_r} G_{ce0} - \gamma_c \gamma_d G_{ce0} G_{cd_r}} \text{ and } \check{P}_{d_t} = \frac{\gamma_d N_0 (\gamma_c G_{cd_r} + G_{ce0})}{G_{d_t d_r} G_{ce0} - \gamma_d \gamma_c G_{ce0} G_{cd_r}}, \quad (16)$$

respectively. From Equation 16, it is obvious that the Problem (P1) is only feasible if $P_{\max C} \geq P_c > 0$ and $P_{\max D} \geq P_{d_t} > 0$. In this work, we assume that $P_{\max C} \geq P_c > 0$ and $P_{\max D} \geq P_{d_t} > 0$ hold. Similarly, the upper bound for \check{P}_{d_t} is given by:

$$\hat{P}_{d_t} = \min \left\{ \frac{1}{G_{d_t e0}} \left(\frac{P_{\max C} G_{ce0}}{\gamma_c} - N_0 \right), P_{\max D} \right\}, \quad (17)$$

which also makes sure that the CUEs transmit power does not exceed $P_{\max C}$.

In order to further simplify Problem (P1), we define $F(w, P_{d_t})$ as:

$$F(w, P_{d_t}) = \frac{1}{P_{d_t}} (R_c + R_d) - w \geq 0, \quad (18)$$

where w is an auxiliary decision variable. Moreover, $F(w, P_{d_t}) \geq 0$ means that:

$$\frac{1}{P_{d_t}} (\log_2 (1 + \text{SINR}_C) + \log_2 (1 + \text{SINR}_{D2D})) \geq w, \quad (19)$$

thus, w can be defined as an adjustable threshold, which DUEs energy efficiency should exceed. According to [50], Problem (P1) can be rewritten as:

$$\max_{\hat{P}_{d_t} \geq P_{d_t} \geq \check{P}_{d_t}} w \quad \text{subject to: } F(w, P_{d_t}) \geq 0. \quad (20)$$

After multiplying P_{d_t} on both sides, Equation 19 can also be written as:

$$F1(w, P_{d_t}) = \frac{1}{P_{d_t}} (\log_2 (1 + \text{SINR}_C) + \log_2 (1 + \text{SINR}_{D2D})) - w P_{d_t}. \quad (21)$$

We present the unique optimal solution for Problem (P1) as:

$$P_{d_t}^* = \begin{cases} \hat{P}_{d_t} & \text{if } 0 < w > \underline{N}, \\ \check{P}_{d_t} & \text{if } w > \bar{N}, \\ \frac{-Y + \sqrt{Y^2 - 4XZ}}{2X} & \text{if } \underline{N} \leq w \leq \bar{N}. \end{cases} \quad (22)$$

The threshold values of \underline{N} and \bar{N} are given by:

$$\underline{N} = \frac{G_{d_t d_r} G_{ce0} (N_0 \gamma_c G_{cd_r} + N_0 G_{ce0})}{(\ln 2) \cdot (X \hat{P}_{d_t}^2 + Y \check{P}_{d_t} + (N_0 \gamma_c G_{cd_r} + N_0 G_{ce0})^2)},$$

$$\bar{N} = \frac{G_{d_t d_r} G_{ce0} (N_0 \gamma_c G_{cd_r} + N_0 G_{ce0})}{(\ln 2) \cdot (X \check{P}_{d_t}^2 + Y \hat{P}_{d_t} + (N_0 \gamma_c G_{cd_r} + N_0 G_{ce0})^2)},$$

where $X = \gamma_c G_{d_t d_r} G_{ce0} G_{cd_r} G_{d_t e0} + (\gamma_c G_{cd_r} G_{d_t e0})^2$, $Y = G_{d_t d_r} G_{ce0} (N_0 \gamma_c G_{cd_r} + N_0 G_{ce0}) + 2 \gamma_c G_{cd_r} G_{d_t e0} (N_0 \gamma_c G_{cd_r} + N_0 G_{ce0})$, and $Z = (N_0 \gamma_c G_{cd_r} + N_0 G_{ce0})^2 - \frac{G_{d_t d_r} G_{ce0} (N_0 \gamma_c G_{cd_r} + N_0 G_{ce0})}{w(\ln 2)}$. Once $P_{d_t}^*$ is obtained, CUEs

can set their transmit power to $P_c^* = \frac{\gamma_c (N_0 + P_{d_t}^* G_{d_t e0})}{G_{ce0}}$. More detailed derivation process for $P_{d_t}^*$ is provided in Appendix.

4 Performance evaluation

4.1 Simulation environment

To evaluate the performance of the proposed scheme, we conducted C-language-based system-level simulations. A

total of 500 simulations are performed, each with a run time of 1,000 secs. We consider a FFR-based cellular network operating in the TDD mode, where N_C CUEs and N_D D2D pairs are randomly deployed in a cell with uniform distribution. The network supports full duplex cellular and half duplex D2D communications, where D2D communications reuse the uplink resource of cellular network. A full buffer traffic model is considered where UEs always have data to transmit.

The shadowing standard deviation of the large-scale fading and path loss exponent is set to 8 and 3.5 dB, respectively [51]. For small-scale multipath fading, a 20-tap typical urban channel model [52] is considered. Modulation and coding scheme (MCS) [53] based adaptive modulation and coding scheme is considered for the system. More detailed simulation parameters are given in Table 2.

To measure the throughput fairness among UEs, we have used Jain's fairness index indicator [54]. Jain's fairness index J can be calculated as:

$$J(x_1, x_2, \dots, x_n) = \frac{(\sum_{i=1}^n x_i)^2}{n \cdot \sum_{i=1}^n x_i^2}, \quad (23)$$

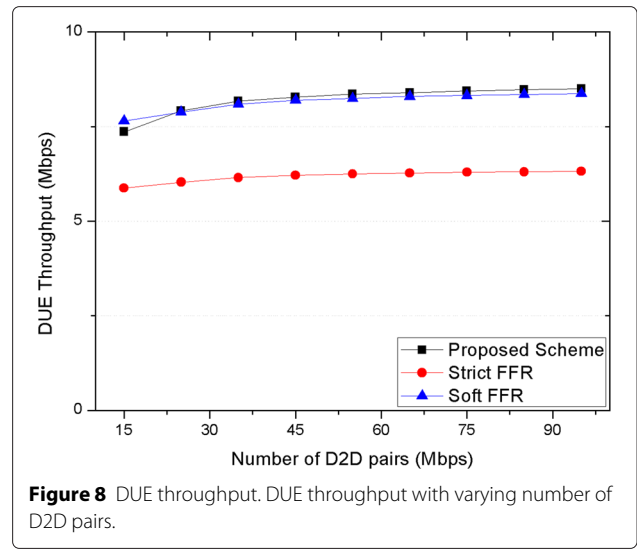
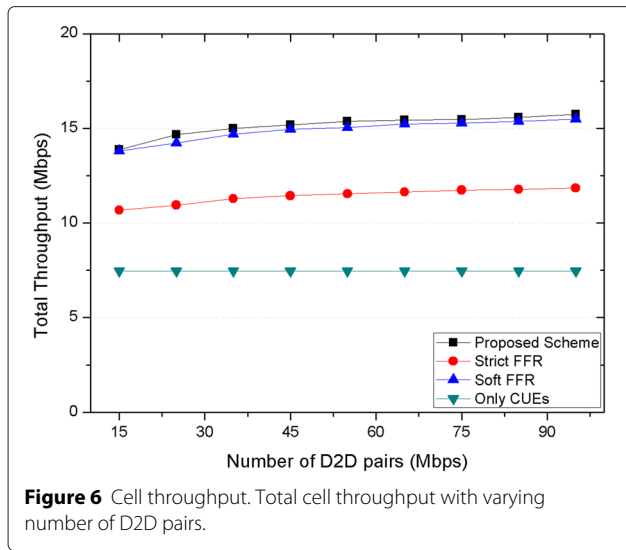
where x_i and n denote the throughput of i th UE and the total number of UEs, respectively. The value of Jain's fairness index ranges from 0 to 1, so that the values closer to 1 represent higher fairness.

4.2 Simulation results

Figure 6 shows the total cell throughput with varying number of D2D pairs. The values of γ_c and γ_d are set to 5 and 0 dB, respectively. The value of N_C in all simulations is set to 30. It can be seen from Figure 6 that

Table 2 Simulation parameters

Parameters	Values
The number of CUEs per cell (N_C)	30
Carrier frequency	2.0 GHz
Uplink bandwidth	5 MHz
Total number of subchannels	25
Maximum UE transmission power (P_{\max})	24 dBm [55]
Subchannel bandwidth	180 KHz
Channel model	The 20-tap typical urban channel model [52]
Path loss exponent (alpha)	3.5
Transmission time interval (TTI)	1 msec
Antenna Gain of eNB	15.0 dB
Antenna Gain of UE	4.0 dB
Noise power density	-174 dBm
Transmission power of eNB	43 dBm [56]



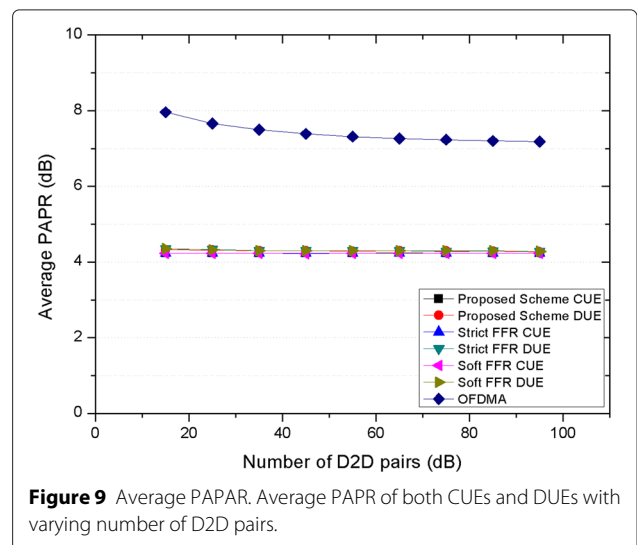
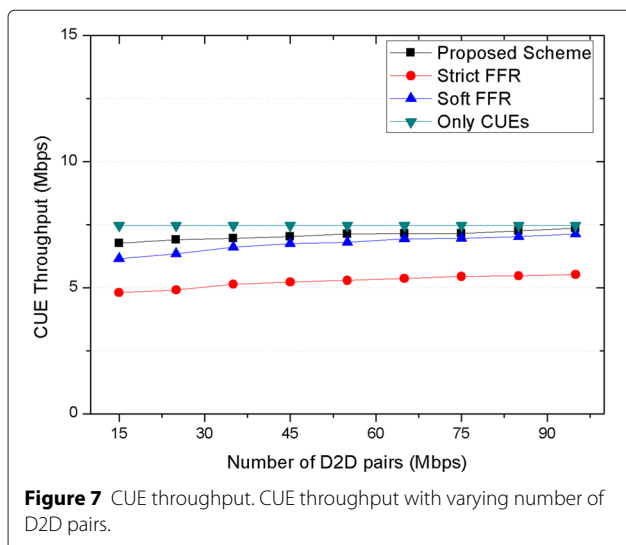
the proposed scheme outperforms both strict FFR and soft FFR schemes. Unlike the proposed scheme, strict FFR does not use any power control mechanism and does not utilize the complete frequency spectrum in each cell. This is why strict FFR results in lower cell throughput. While soft FFR utilizes the entire frequency spectrum in each cell, it also does not use any power control schemes. The proposed scheme outperforms both soft and strict FFR schemes because it wisely considers the spatial distance between resource sharing pairs while allocating the resources. Only CUEs scheme (conventional cellular setup with no D2D communications) performs worst as expected.

Figure 7 shows the total throughput of CUEs with varying number of D2D pairs. It is obvious from Figure 7 that the proposed scheme achieves higher throughput

for CUEs in comparison with strict and soft FFR schemes. However, the only-CUEs scheme outperforms all other schemes because there is no D2D communications and CUEs do not observe any interference at all.

The total throughput of DUEs with varying number of D2D pairs has been shown in Figure 8. As shown in Figure 8, the proposed scheme and soft FFR outperform strict FFR in terms of DUE throughput. It can also be observed that the proposed scheme performs slightly better than the soft FFR scheme.

The average PAPR of both CUEs and DUEs with different numbers of D2D pairs has been shown in Figure 9. In order to assess the performance of the proposed scheme in terms of PAPR, we have compared our results with an OFDMA-based resource allocation



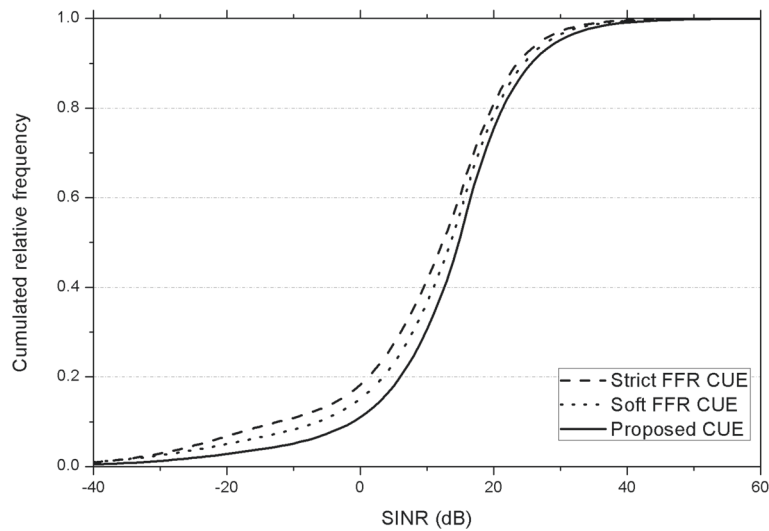


Figure 10 CRF of CUEs' SINR. Cumulative relative frequency (CRF) of CUEs' SINR for $N_C = 30$ and $N_D = 25$.

scheme [18]. Recall from our earlier discussion that due to the sequential transmission of subcarriers in SC-FDMA, it has low PAPR than OFDMA signals. This is one of the main reasons that LTE-A has opted SC-FDMA for uplink transmission. Our results shows that UEs in the proposed scheme have an advantage of low transmission power compared to the previously proposed OFDMA-based D2D communications schemes. However, the performance of the proposed scheme in terms of PAPR is almost similar to strict and soft FFR schemes.

For $N_C = 30$ and $N_D = 25$, Figures 10 and 11 depict the cumulative relative frequency (CRF) of SINR.

Figure 10 shows that our proposed scheme outperforms both strict FFR and soft FFR schemes. Similarly, Figure 11 shows that the SINR of DUEs in the proposed scheme is better than those in both strict FFR and soft FFR schemes. It is because, in order to satisfy the pre-specified minimum SINR, our proposed scheme dynamically controls the transmit power of DUEs in every TTI. In proposed scheme, the percentage of CUEs achieving their target SINR is greater than both strict and soft FFR schemes. Except the strict FFR scheme, about 20% of the DUEs has experienced SINR lower than 12 dB in the soft FFR and our proposed schemes. The reason is that DUEs reduce their transmit power in order to satisfy the minimum

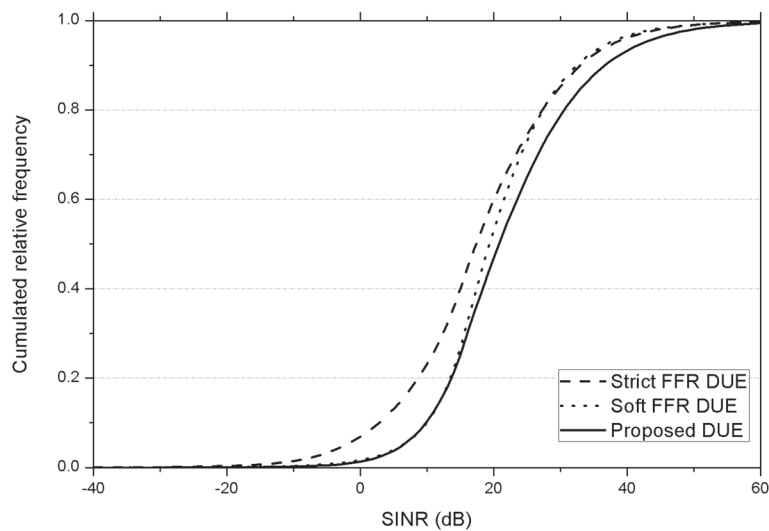
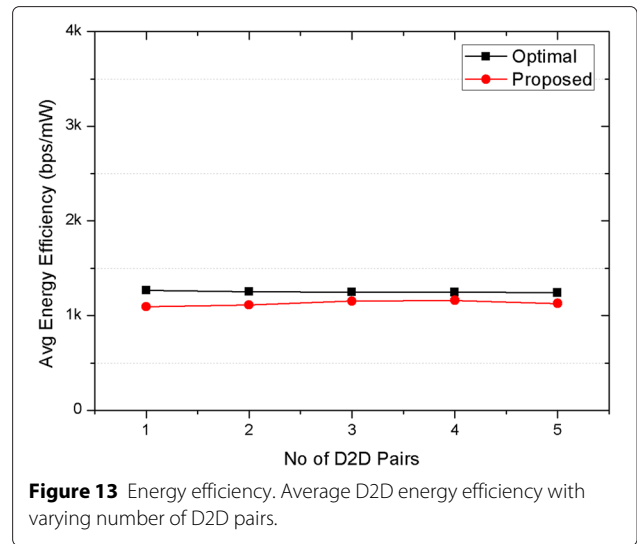
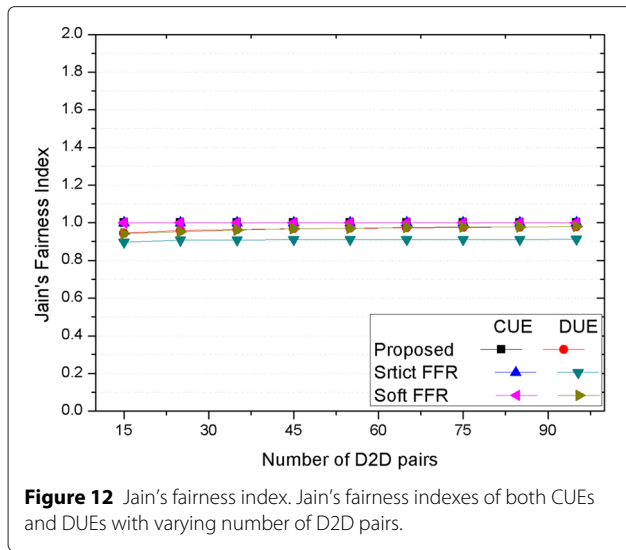


Figure 11 CRF of DUEs' SINR. Cumulative relative frequency (CRF) of DUEs' SINR for $N_C = 30$ and $N_D = 25$.



SINR of CUEs. However, for the rest of 80% DUEs, the proposed scheme outperforms both strict and soft FFR schemes.

Figure 12 shows Jain's fairness indexes of both CUEs and DUEs with varying number of D2D pairs. For CUEs, the value of Jain's fairness index is almost the same in all three schemes. However, for DUEs, the proposed scheme and soft FFR scheme outperform the strict FFR scheme. It can also be seen that the proposed scheme does better than soft FFR in terms of fairness.

To evaluate the energy efficiency, we consider five CUEs and five DUEs randomly deployed in a cell such that the number of sub-channels is also limited to five. Figure 13 depicts the D2D energy efficiency of both optimal and proposed schemes. It can be observed from the figure that in terms of energy efficiency, the proposed scheme performs near to optimal. The slight difference between the optimal and proposed scheme is due to the fact that the proposed scheme does not use any lower bound for its DUEs transmit power. Nevertheless, in a densely populated network, because of severe interference between resource sharing pairs, it is unlikely for D2D users to achieve their target data rates while transmitting with minimum power. In other words, to achieve their target data rates in a densely deployed D2D network, DUEs have to transmit with transmission power higher than the lower bound derived in the optimal solution.

5 Conclusions

D2D communications reusing the uplink resources of cellular users can significantly improve the network capacity. In this paper, we have proposed an FFR-based resource allocation scheme for D2D communications using SC-FDMA and PF scheduler as a multiple access technique and scheduling algorithm, respectively. The proposed scheme also adopts an intelligent power control mechanism for D2D users which does not only guarantees a certain SINR for CUEs but also provides equal opportunity for DUEs to maintain a minimum SINR value. The simulation result shows that the proposed scheme alleviates the throughput of all UEs and improves the overall cell capacity. Furthermore, due to the use of SC-FDMA, the proposed scheme also alleviates the high PAPR problem. As a future research work, this approach can be extended by considering the reuse of both uplink and downlink radio resources for D2D communications.

Appendix

In order to maximize the energy efficiency the CUEs should transmit with the power P_c that guarantees its minimum throughput which according to Equation 15 is $P_c = \frac{\gamma_c(N_0 + P_{d_t} G_{d_t e 0})}{G_{c e 0}}$. By using this setting for P_c in $F1(w, P_{d_t})$ (Equation 21), its first-order derivative $\frac{\partial F1(w, P_{d_t})}{\partial P_{d_t}}$ becomes:

$$\frac{\partial F1(w, P_{d_t})}{\partial P_{d_t}} = \frac{G_{d_t d_r} G_{c e 0} (N_0 \gamma_c G_{c d_r} + N_0 G_{c e 0})}{(\ln 2) \cdot \left((\gamma_c G_{c d_r} G_{d_t e 0} P_{d_t} + (\gamma_c G_{c d_r} + G_{c e 0}) N_0 \right)^2 \left(\frac{G_{d_t d_r} G_{c e 0} P_{d_t}}{\gamma_c G_{c d_r} G_{c e 0} P_{d_t} + (\gamma_c G_{c d_r} + G_{c e 0}) N_0} + 1 \right)} - w, \quad (24)$$

which shows that $F1(w, P_{d_t})$ is concave in P_{d_t} because the first-order derivative is strictly decreasing in P_{d_t} . Hence Problem (P1) is a convex optimization problem. The convex nature of the problem allows us to use the Karush-Kuhn-Tucker (KKT) conditions to solve the problem. The first two cases in Equation 22 are obtained by taking \hat{P}_{d_t} and \check{P}_{d_t} into account, which limits the DUEs transmit power (Equations 16 and 17). Moreover, we obtain the quadratic equation $XP_{d_t}^2 + YP_{d_t} + Z = 0$ by setting $\frac{\partial F1(w, P_{d_t})}{\partial P_{d_t}} = 0$, whose positive root, if it exists, corresponds to the third case in Equation 22. The threshold \bar{N} and \underline{N} (for w) are obtained by setting $P_{d_t}^* = \check{P}_{d_t}$ and \hat{P}_{d_t} , respectively.

Competing interests

The authors declare that they have no competing interests.

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