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Secrecy analysis of cognitive radio network with MS-GSC/MRC scheme

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Abstract

We propose and analyze minimum selection-generalized selection combining (MS-GSC) at secondary receiver (SR) with maximal ratio combining at eavesdropper (ER) to enhance data security at physical layer. We consider an underlay cognitive radio network (CRN) where SR and ER are equipped with multiple antennas, and secondary transmitter (ST) has single antenna with a primary user. Passive eavesdropping is also taken into account. This work is aimed to find the effect of MS-GSC diversity technique on secrecy outage probability (SOP). We derive a closed-form expressions for the exact and asymptotic SOP. Our results show a positive impact on SOP with an increase in diversity branches and also reveal the effect of a primary user on secondary network.

Keywords: Cognitive radio network, Minimum selection-generalized selection combining (MS-GSC), Outage probability, Physical layer security

1 Introduction

In wireless communication systems, an eavesdropper intercepts transmission due to the broadcast nature of wireless links. So, the security of data transmission in these networks is becoming more critical than ever [1, 2]. Traditionally, cryptographic techniques are used to secure data at the upper layer of protocol stack using public and private key variations. In underlay cognitive radio networks (CRN), the primary users (PU) and the secondary users (SU) transmit concurrently in the same band of frequency [3, 4]. The protection and security of the broadcast channel in such complex environments against eavesdropping is a very difficult task. The conventional cryptographic authentication become very expensive and less effective because of the open nature of these broadcasting channels [5, 6]. Therefore, research efforts have been devoted to physical layer security, which exploits the characteristics of wireless channel (e.g., thermal noise and fading) to secure the communication at physical layer [7]. The fundamental concept of physical layer security is to enhance the gain of legitimate receiver's main channel in comparison to the eavesdropper's channel to attain perfect secrecy. With the advancement in multiple antenna

techniques, security improvement in wiretap CRN channels has addressed from information-theoretic perspective [8–11], where the secondary transmitter (ST), the secondary user receiver (SR), and the eavesdropper (ER) all are equipped with multiple antennas.

In the last few years, lots of work had been done to enhance the security of physical layer in CRN. In [12, 13], secrecy performance of single input multiple output (SIMO) CRN with maximal ratio combining (MRC) diversity technique was studied. Secrecy outage probability (SOP) for transmit antenna selection (TAS)/MRC in multiple input multiple output (MIMO) CRN has been investigated in [14], and an underlay MIMO CRN with a pair of primary nodes and secondary nodes and an eavesdropper was considered in [15] where the secondary transmitter was powered by the renewable energy harvested from the primary transmitter in order to improve both energy efficiency and spectral efficiency. SOP of an underlay cognitive decode-and-forward relay network over independent but not necessarily identical distributed (i.n.i.d) Nakagami-m fading channels was investigated, and optimal relay selection (ORS) and suboptimal relay selection (SRS) schemes, and multiple relay combining scheme were considered in [16]. In [17], a hybrid visible light communication radio frequency (RF) system with legitimate receiver and an eavesdropper was considered where legitimate receiver can harvest energy from the light emitted

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by light-emitting diodes, and exact and asymptotic SOP was derived by using stochastic geometry method. The closed-form expressions for SOP and non-zero secrecy capacity for underlay CR unit over Nakagami- m fading were investigated in [18], and the secrecy outage performance of PU system in the presence of the eavesdropping and interfering of SU was analyzed in [19]. Generalized-selection combining (GSC) is a hybrid technique that overcomes the limitation of MRC and selection combining (SC). In [20–22], a detailed study has been done on GSC. GSC uses a fixed number of the best branches (M_C) of all available ones (M). In GSC, there is no need of process all the paths that reduce the hardware complexity of receiver, but all MRC branches remain active during the reception of the data, which increases consumption of computational power [20]. The secrecy analysis for SIMO wiretap CRN with GSC over Nakagami- m fading channels was done in [22]. TAS with GSC over Rayleigh fading and Nakagami- m fading channels were applied in [23–26]. In [26], TAS/GSC for cognitive decode-and-forward relaying in Nakagami- m fading channels was considered and closed-form expression for ergodic capacity was derived. The power saving implementation of GSC called minimum selection GSC (MS-GSC) had proposed in [27].

In this paper, we consider a communication scenario where MS-GSC diversity combining technique is adopted by the SR considering the complexity and energy dissipation, and in order to maximize, its instantaneous SNR at ER MRC technique is applied. In fact, MS-GSC is a more general diversity combining than GSC. The basic idea of MS-GSC is that the minimum number of diversity branches are selected such that their combined signal-to-noise ratio (SNR), and not the individual branch SNR, is above a given threshold [28].

Our contributions are as follows:

1. We examine the secrecy performance of an underlay CRN with MS-GSC applied at legitimate receiver and MRC at eavesdropper over Rayleigh fading environment and derived closed-form expression for SOP.
2. In [2], interference power constraint was considered at primary user (PU). Both SR and ER were equipped with multiple antennas, and selection combining (SC) was applied. SC select only one antenna with the highest SNR among available ones which neglect diversity phenomena. In comparison to [2], interference power constraint, more generalized, and computational power saving system is considered in this paper.
3. In [12], MRC scheme was used over Rayleigh fading channels at both secondary receiver SR and eavesdropper. We apply MS-GSC at SR and MRC at

eavesdropper and SC ($M_C = 1$), MRC ($M_C = M$), and GSC ($M_C \leq M$) are the special cases of MS-GSC.

The organization of the remaining paper is as follows: Section 2 explains the proposed system model. Furthermore, the working of MS-GSC is also explained in Section 2. Secrecy outage probability is calculated in Section 3. Section 4 gives the interpretation of numerical result.

2 System model

Here, we consider an underlay wiretap CRN composed of a primary user (PU), secondary user transmitter as Alice, legitimate receiver as B, and an eavesdropper as ER. The primary user and Alice are consist of a single antenna whereas B and the ER are assumed to have multiple antennas M . Here, we assume that the confidential messages are being transmitted from Alice to B in the presence of ER, and the ER want to listen their communication. In order to have reliable communication, the interference power at PU from Alice should be less than the peak interference power threshold.

The primary and secondary channels are experiencing i.i.d Rayleigh fading where channel gains of the main channel $\{h_{Bt}\}_{t=1}^M$, eavesdropper's channel $\{h_{Es}\}_{s=1}^M$, and primary channel h_p are complex Gaussian random variables with zero mean and variances Ω_1, Ω_2 , and Ω_0 , respectively. We are considering passive eavesdropping, i.e., the channel state information (CSI) of the eavesdropper is also available at Alice. The channel gain and variance of the primary users are h_p and Ω_0 , respectively. The main channels (Alice to B) and eavesdropper channels (Alice to ER) are independent of each other. Here, the MS-GSC diversity scheme is applied at B, and MRC diversity technique is applied at ER. In this paper, we assume that the global CSI of the main links, the eavesdropper's links, and primary link is available for evaluating the secrecy rate in the information receiver, which is a common assumption in the literature on physical layer security. Information on the PU's channels can be obtained for the cases in which the PUs are cooperative in the network and their transmissions can be monitored. This is applicable for those networks that combining multicast and unicast transmissions, in which terminals play dual roles as legitimate receivers for some signals and eavesdroppers for others. In practice, this information can be generated by using reverse training, where the primary user transmits training signal to Alice such that by invoking the principle of reciprocity, Alice can estimate the primary users CSI.

The interference received at secondary network from primary network is considered to be a complex Gaussian random variable under an assumption that the primary signal may be generated by the random Gaussian codebook. Moreover, the thermal noise at secondary nodes

is also complex Gaussian distributed. Thus, the interference plus noise at secondary nodes (B and eavesdroppers) can be modeled as a complex Gaussian random variable with zero mean and variance N_0 . It indicates that the influence of primary interference at secondary nodes is already assumed in the variance N_0 . Thus, the effect of interference is adjusted in the noise statistics at B and ER [29, 30]. The instantaneous SNR of the main channels and eavesdropper's channel is given by

$$\Psi_M = \max_{t=1,\dots,M} \frac{P_{av}}{N_0} |h_{B_t}|^2, \Psi_E = \max_{s=1,\dots,M} \frac{P_{av}}{N_0} |h_{E_s}|^2 \quad (1)$$

where P_{av} is Alice's transmitted power.

2.1 Working of MS-GSC diversity technique

The receiver with MS-GSC diversity schemes chooses the minimum number of best available antennas in such a way that the combined SNR Ψ_{com} is always more than the threshold SNR Ψ_t [22]. Mathematically, the working of MS-GSC can be summarized as

$$\Psi_{com} = \begin{cases} \Psi(1) & \text{iff } \Psi_1 \geq \Psi_t; \\ \Psi(1) + \Psi(2) & \text{iff } \Psi(1) < \Psi_t \text{ \& } \Psi(1) + \Psi(2) \geq \Psi_t; \\ \vdots & \\ \sum_{k=1}^r \Psi(k) & \text{iff } \sum_{k=1}^{r-1} \Psi(k) < \Psi_t \text{ \& } \sum_{k=1}^r \Psi(k) \geq \Psi_t; \\ \vdots & \\ \sum_{k=1}^{M_c} \Psi(k) & \text{iff } \sum_{k=1}^{r-1} \Psi(k) < \Psi_t. \end{cases} \quad (2)$$

where $\psi_1, \psi_2, \dots, \psi_k$ are the SNR of path 1, path 2 path k , respectively.

GSC uses M_c antennas among available M antennas such that $M_c < M$. In comparison to GSC, MS-GSC needs less computing power, and on average, less number of MRC branches is active during the reception of data which save processing power. Let N_b be the active number of MRC branches, which takes the value from 1 to M_c . $N_b = 1$ if $\Psi_1 > \Psi_t$, $N_b = r$, $2 \leq r \leq M_c - 1$ if and only if $\sum_{k=1}^{r-1} \Psi_k < \Psi_t$ and $\sum_{k=1}^r \Psi_k \geq \Psi_t$, and $N_b = M_c$ if $\sum_{k=1}^{M_c-1} \Psi_k < \Psi_t$.

3 Secrecy outage probability

The secrecy capacity analysis can help us to determine how secure a CRN is and whether we need more security mechanisms to protect against the potential attacks in the CRNs. The maximum achievable rate is named as secrecy capacity and given by ([2], Eq.2). The secrecy capacity of CRN consists an antenna at *Alice* and multiple antennas at B and ER can be defined as,

$$S_c = \begin{cases} M_c - E_e = \log_2 \left(\frac{1+\Psi_M}{1+\Psi_E} \right) & \text{if } \Psi_M > \Psi_E, \\ 0 & \text{if } \Psi_M \leq \Psi_E, \end{cases} \quad (3)$$

where $M_c = \log_2(1 + \Psi_M)$ is the capacity of channels between Alice and B and $E_e = \log_2(1 + \Psi_E)$ is the capacity of channels between *Alice* and ER . S_c in (4) can be rewritten as

$$S_c = \log_2 \left(\frac{1 + \Psi_M}{1 + \Psi_E} \right) < R_s \quad (4)$$

which is analogous to

$$\epsilon(\Psi_E) = 2^{R_s}(1 + \Psi_E) - 1 > \Psi_M. \quad (5)$$

In passive eavesdropping, excellent secrecy is possible if and only if $R_s \leq S_c$; otherwise, information-theoretic security is compromised. SOP is a probability that secrecy capacity S_c falls under the output threshold R_s [2] and is given as

$$P_{out} = P_r(S_c < R_s) = P_r(\Psi_M \leq \Psi_E) + P_r(\Psi_M > \Psi_E) P_r(S_c < R_s | \Psi_M > \Psi_E) \quad (6)$$

which can be simplified to (7)

$$P_{out} = \int_0^\infty \int_0^\infty F_{\Psi_M|Y=y}(\epsilon(\Psi_E)) f_{\Psi_E|Y=y}(\Psi_E) f_Y(y) d\Psi_E dy \quad (7)$$

where $Y = |h_p|^2$ is the channel gain from Alice to PU, $f_Y(y)$ is the probability density function (PDF) of Y , $f_{\Psi_E|Y=y}$ is the PDF of Ψ_E conditioned on Y , and $F_{\Psi_M|Y=y}(\epsilon(\Psi_E))$ is the cumulative distribution function (CDF) of Ψ_M conditioned on Y .

In underlay cognitive radio transmission, for reliable communication, *Alice's* transmitted power P_{av} should be less than the peak interference power threshold. So, *Alice* is a power-limited transmitter with a maximum transmit power which is P_T . The transmitted power of Alice is constrained by P_T at Alice and peak interference power P_I at the primary user and given by

$$P_{av} = \min \left(\frac{P_I}{h_p}, P_T \right) \quad (8)$$

Based on (8) instantaneous SNR at secondary receiver, B and ER are expressed as

$$\psi_M = \min \left(\frac{\psi_p}{Y}, \psi_0 \right) X_M \quad \psi_E = \min \left(\frac{\psi_p}{Y}, \psi_0 \right) X_E \quad (9)$$

where $\psi_p = \frac{P_I}{N_0}$, $\psi_0 = \frac{P_T}{N_0}$, $X_M = \max_{t=1,\dots,M} |h_{B_t}|^2$ and $X_E = \max_{s=1,\dots,M} |h_{E_s}|^2$.

For ease of interpretation, we have $\Psi_1 = \Omega_1 \Psi_0 = \psi_p \frac{\Omega_1}{\sigma}$ be the average SNR of the main channel, $\Psi_2 = \Omega_2 \Psi_0 = \psi_p \frac{\Omega_2}{\sigma}$ be the average SNR of the ED's channel, and $\sigma = \frac{P_I}{P_T}$. The CDF for MS-GSC scheme for Rayleigh fading is given by ([22], Eq. 24)

$$P(x_1) = \begin{cases} P_{\gamma_{M_c}}(x_1), & 0 \leq x_1 < \Psi_t; \\ P_{\gamma_1}(x_1) - P_{\gamma_1}(\Psi_t) + P_{\gamma_{M_c}}(\Psi_t) \\ + \sum_{w=2}^{M_c} \left(\int_{\frac{w-1}{w}\Psi_t}^{\frac{w-1}{w}x_1} \int_{\Psi_t-y}^y p_{\Psi(w)}, \gamma_{w-1}(z, y) dz dy \right. \\ \left. + \int_{\frac{w-1}{w}x_1}^{\Psi_t} \int_{\Psi_t-y}^{x_1-y} p_{\Psi(w)}, \gamma_{w-1}(z, y) dz dy \right), & \Psi_t \leq x_1 < \frac{M_c}{M_c-1} \Psi_t; \\ \vdots \\ P_{\gamma_1}(x_1) - P_{\gamma_1}(\Psi_t) + P_{\gamma_{M_c}}(\Psi_t) \\ + \sum_{v=2}^{M_c} \left(\int_{\frac{v-1}{v}\Psi_t}^{\frac{v-1}{v}x_1} \int_{\Psi_t-y}^y p_{\Psi(w)}, \gamma_{w-1}(z, y) dz dy \right. \\ \left. + \int_{\frac{v-1}{v}x_1}^{\Psi_t} \int_{\Psi_t-y}^{x_1-y} p_{\Psi(w)}, \gamma_{w-1}(z, y) dz dy \right), & \frac{v+1}{v} \Psi_t \leq x_1 < \frac{v}{v-1} \Psi_t; \\ \vdots \\ P_{\gamma_1}(x_1), & 2\Psi_t < x_1 \end{cases} \quad (10)$$

The CDF $P_{\Psi_w}(\cdot)$ is given by ([22], Eq. 17)

$$P_{\gamma_w}(x_1) = \frac{M!}{(M-w)!w!} \left\{ 1 - e^{-\frac{x_1}{\Psi_1}} \sum_{k=0}^{w-1} \frac{1}{k!} \left(\frac{x_1}{\Psi_1} \right)^k \right. \\ + \sum_{v=1}^{M-w} (-1)^{w+v-1} \frac{(M-w)!}{(M-w-v)!} \left(\frac{w}{v} \right)^{w-1} \\ \times \left[\left(1 + \frac{v}{w} \right)^{-1} \left[1 - e^{-\left(1 + \frac{v}{w} \right) \frac{x_1}{\Psi_1}} \right] \right. \\ \left. \left. - \sum_{m=0}^{w-2} \left(\frac{-v}{w} \right)^m \left(1 - e^{-\frac{x_1}{\Psi_1}} \sum_{k=0}^m \frac{1}{k!} \left(\frac{x_1}{\Psi_1} \right)^k \right) \right] \right\} \quad (11)$$

where $x_1 = 2^{Rs}(1+x)-1$ and the PDF of MRC $[p_{\gamma_M}(\cdot)]$ is given by

$$p_{\Psi_M}(x) = e^{-\frac{x}{\Psi_2}} \frac{x^{M-1}}{\Psi_2^M (M-1)!} \quad (12)$$

where Ψ_2 is the average SNR of the ER's channels. The PDF of Y is given by ([2], Eq. 22)

$$f_R(r) = \sum_{h=0}^{N-1} \frac{N}{\Omega_0} (-1)^h e^{-\frac{(h+1)r}{\Omega_0}}, \quad y \geq 0 \quad (13)$$

for $N = 1$

$$f_R(r) = \frac{1}{\Omega_0} e^{-\frac{r}{\Omega_0}}, \quad r \geq 0 \quad (14)$$

Using all the above mentioned equations in (7), SOP for proposed system model is given by (15).

$$P_{out} = \begin{cases} P_{outA} & 0 \leq x_1 < \Psi_t; \\ P_{outB} & \Psi_t \leq x_1 < \frac{M_c}{M_c-1} \Psi_t \\ \vdots \\ P_{outC} & \frac{v+1}{v} \Psi_t \leq x_1 < \frac{v}{v-1} \Psi_t \\ \vdots \\ P_{outD} & 2\Psi_t < x_1 \end{cases} \quad (15)$$

The detail expansion of the above equation is as follows:

$$p_{outA} = \frac{M!}{M_c!(M-M_c)!} \left\{ \left(1 - e^{-\frac{x}{\Omega_0}} \right) \left(1 - \sum_{k=0}^{M_c-1} \sum_{n=0}^k \mu_1 \right) \right. \\ e^{-\left(\frac{2^{Rs}-1}{\Psi_1} \right)} + \sum_{v=1}^{M-M_c} C_2 C_3 - \sum_{v=1}^{M-M_c} \sum_{m=0}^{M_c-2} C_2 C_4 - \sum_{v=1}^{M-M_c} C_2 C_4 \beta_1 \\ e^{-\left(\frac{(2^{Rs}-1)(1+\frac{v}{M_c})}{\Psi_1} \right)} + \sum_{v=1}^{M-M_c} \sum_{m=0}^{M_c-2} \sum_{k=0}^m \sum_{n=0}^k C_2 C_4 \mu_1 \\ e^{-\left(\frac{2^{Rs}-1}{\Psi_1} \right)} + e^{-\frac{x}{\Omega_0}} - \sum_{k=0}^{M_c-1} \sum_{n=0}^k \sum_{p=0}^{k-n} Q_1 Q_2 e^{-\sigma \left(\frac{2^{Rs}-1}{\sigma \Psi_1} + \frac{1}{\Omega_0} \right)} \\ + \sum_{v=1}^{M-M_c} C_2 C_3 e^{-\left(\frac{x}{\Omega_0} \right)} - \sum_{v=1}^{M-M_c} C_2 C_3 \beta_2 \\ e^{-\sigma \left(\left(1 + \frac{M}{M_c} \right) \left(\frac{2^{Rs}-1}{\sigma \Psi_1} \right) + \frac{1}{\Omega_0} \right)} - \sum_{v=1}^{M-M_c} \sum_{m=0}^{M_c-2} C_2 C_4 e^{-\left(\frac{x}{\Omega_0} \right)} \\ \left. + \sum_{v=1}^{M-M_c} \sum_{m=0}^{M_c-2} \sum_{k=0}^m \sum_{n=0}^k \sum_{p=0}^{k-n} C_2 C_4 Q_1 Q_2 e^{-\sigma \left(\frac{2^{Rs}-1}{\sigma \Psi_1} + \frac{1}{\Omega_0} \right)} \right\} \quad (16)$$

where

$$\mu_1 = \frac{1}{(M-1)!k! \Psi_1^k} \binom{k}{n} (2^{Rs}-1)^{k-n} (2^{Rs})^n \frac{(n+M-1)!}{(\Psi_2)^M \left(\frac{2^{Rs}}{\Psi_1} + \frac{1}{\Psi_2} \right)^{n+M}} \\ C_2 = \sum_{v=1}^{M-M_c} (-1)^{M_c+v-1} \frac{(M-M_c)!}{(M-M_c-v)!} \left(\frac{M_c}{v} \right)^{M_c-1} \\ C_3 = \left(1 + \frac{v}{M_c} \right)^{-1}, \quad C_4 = \left(\frac{-v}{M_c} \right)^m \\ \beta_1 = \frac{1}{\Psi_2^M \left(\frac{(1+\frac{v}{M_c})2^{Rs}}{\Psi_1} + \frac{1}{\Psi_2} \right)^M} \\ Q_1 = \frac{1}{k! (\sigma \Psi_1)^k} (2^{Rs}-1)^{k-n} (2^{Rs})^n \binom{k}{n} \frac{(n+M-1)!}{\left(\frac{2^{Rs}}{\sigma \Psi_1} + \frac{1}{\Psi_2} \right)^{n+M}} \\ Q_2 = \frac{(\sigma)^p (k-n)!}{(\sigma \Psi_2)^M (M-1)! p! \left(\frac{2^{Rs}-1}{\sigma \Psi_1} + \frac{1}{\Omega_0} \right)^{k-n-p+1}} \\ \beta_2 = \frac{1}{(\sigma \Psi_2)^M \left(\frac{(1+\frac{v}{M_c})2^{Rs}}{\sigma \Psi_1} + \frac{1}{\sigma \Psi_2} \right)^M} \\ \times \frac{1}{\left(\left(1 + \frac{M}{M_c} \right) \left(\frac{2^{Rs}-1}{\sigma \Psi_1} \right) + \left(\frac{1}{\Omega_0} \right) \right)}$$

$$P_{out_B} = P_{out_D} - P_{out_1}(\Psi_t) + P_{out_{M_c}}(\Psi_t) + \sum_{w=2}^{M_c} P_{out_M} \quad (17)$$

$$P_{out_1}(\Psi_t) = \sum_{v=0}^M \binom{M}{v} (-1)^v \left[e^{-\frac{v\Psi_t}{\Psi_1}} \left(1 - e^{-\frac{\sigma}{\Omega_0}} \right) + \frac{e^{-\sigma \left(\frac{v\Psi_t}{\Psi_1} + \frac{\sigma}{\Omega_0} \right)}}{\left(\frac{v\Psi_t}{\sigma\Psi_1} + \frac{1}{\Omega_0} \right)} \right] \quad (18)$$

$$P_{out_M} = I_1 + I_2 \quad (19)$$

$$I_1 = \int_0^\sigma \int_0^\infty \left[\int_{\frac{w-1}{w}\Psi_t}^{\frac{y}{w-1}} P_{\Psi(w), \Upsilon_{w-1}}(z, y) dz dy + \int_{\frac{w-1}{w}x_1}^{\Psi_t} P_{\Psi(w), \Upsilon_{w-1}}(z, y) dz dy \right] e^{-\frac{x}{\Psi_2}} \frac{x^{M-1}}{\Psi_2^M (M-1)!} \frac{1}{\Omega_0} e^{-\frac{r}{\Omega_0}} dx dr \quad (20)$$

$$I_2 = \int_0^\sigma \int_0^\infty \left[\int_{\frac{w-1}{w}\Psi_t}^{\frac{y}{w-1}} P_{\Psi(w), \Upsilon_{w-1}}(z, y) dz dy + \int_{\frac{w-1}{w}x_1}^{\Psi_t} P_{\Psi(w), \Upsilon_{w-1}}(z, y) dz dy \right] e^{-\frac{x}{\Psi_2}} \frac{x^{M-1}}{\Psi_2^M (M-1)!} \frac{1}{\Omega_0} e^{-\frac{r}{\Omega_0}} dx dr \quad (21)$$

$$P_{out_{M_c}}(\Psi_t) = \left(1 - e^{-\frac{\sigma}{\Omega_0}} \right) \frac{M!}{(M - M_c)! M_c!} \left\{ 1 - e^{-\frac{\Psi_t}{\Psi_1}} \sum_{k=0}^{M_c-1} \frac{1}{k!} \left(\frac{\Psi_t}{\Psi_1} \right)^k + \sum_{v=1}^{M-M_c} C_2 \left[C_3 \left(1 - e^{-\left(1 + \frac{v}{M_c} \right) \frac{\Psi_t}{\Psi_1}} \right) - \sum_{m=0}^{M_c-2} C_4 \left(1 - e^{-\frac{\Psi_t}{\Psi_1}} \sum_{k=0}^m \frac{1}{k!} \left(\frac{\Psi_t}{\Psi_1} \right)^k \right) \right] \right\} \\ + \frac{M!}{(M - M_c)! M_c!} \left[e^{-\frac{\sigma}{\Omega_0}} \left(1 + \sum_{v=1}^{M-M_c} C_2 C_3 - \sum_{v=1}^{M-M_c} \sum_{m=0}^{M_c-2} C_2 C_4 \right) \right] \quad (22)$$

$$- \frac{1}{\Omega_0} \sum_{k=0}^{M_c-1} \sum_{l=0}^k \frac{1}{k!} \left(\frac{\Psi_t}{\sigma\Psi_1} \right)^k e^{-\left(\frac{\Psi_t}{\Psi_1} + \frac{\sigma}{\Omega_0} \right) k!} \frac{(\sigma)^l}{\left[\frac{\Psi_t}{\sigma\Psi_1} + \frac{1}{\Omega_0} \right]^{k-l+1}} \left(1 - \sum_{v=1}^{M-M_c} \sum_{m=0}^{M_c-2} C_2 C_4 \right) - \frac{1}{\Omega_0} \sum_{v=1}^{M-M_c} C_2 C_3 \frac{e^{-\left(\frac{\Psi_t}{\Psi_1} + \frac{\sigma}{\Omega_0} \right)}}{\frac{\Psi_t}{\sigma\Psi_1} + \frac{\sigma}{\Omega_0}} \\ P_{out_D} = \left(1 - e^{-\frac{\sigma}{\Omega_0}} \right) \sum_{v=0}^M \binom{M}{v} (-1)^v \frac{e^{-\left(\frac{v(2^{Rs}-1)}{\Psi_1} \right)}}{(\Psi_2)^M \left(\frac{v2^{Rs}}{\sigma\Psi_1} + \frac{1}{\sigma\Psi_2} \right)^M} + \sum_{v=0}^M \binom{M}{v} (-1)^v \left(\frac{1}{(\sigma\Psi_2)^M \left(\frac{v2^{Rs}}{\sigma\Psi_1} + \frac{1}{\sigma\Psi_2} \right)^M} e^{-\left(\frac{v(2^{Rs}-1)}{\Psi_1} + \frac{\sigma}{\Omega_0} \right)} \right) \quad (23)$$

$$P_{out_C} = P_{out_D} - P_{out_1}(\Psi_t) + P_{out_v}(\Psi_t) + \sum_{w=2}^v P_{out_M}$$

$$P_{out_v}(\Psi_t) = \left(1 - e^{-\frac{\sigma}{\Omega_0}} \right) \frac{M!}{(M - v)! v!} \left\{ 1 - e^{-\frac{\Psi_t}{\Psi_1}} \sum_{k=0}^{v-1} \frac{1}{k!} \left(\frac{\Psi_t}{\Psi_1} \right)^k + \sum_{v=1}^{M-v} (-1)^{2v-1} \frac{(M-v)!}{(M-v-v)! v!} \times \left[\frac{1}{2} \left[1 - e^{-\frac{2\Psi_t}{\Psi_1}} \right] \right. \right. \\ \left. \left. - \sum_{m=0}^{v-2} (-1)^m \left(1 - e^{-\frac{\Psi_t}{\Psi_1}} \sum_{k=0}^m \frac{1}{k!} \left(\frac{\Psi_t}{\Psi_1} \right)^k \right) \right] \right\} + \frac{M!}{(M-v)! v!} \left[e^{-\frac{\sigma}{\Omega_0}} \left(1 + \sum_{v=1}^{M-v} \frac{1}{2} K_1 - \sum_{v=1}^{M-v} \sum_{m=0}^{v-2} K_1 (-1)^m \right) \right] \quad (24) \\ - \frac{1}{\Omega_0} \sum_{k=0}^{v-1} \sum_{l=0}^k \frac{1}{k!} \left(\frac{\Psi_t}{\sigma\Psi_1} \right)^k e^{-\left(\frac{\Psi_t}{\Psi_1} + \frac{\sigma}{\Omega_0} \right) k!} \frac{(\sigma)^l}{\left[\frac{\Psi_t}{\sigma\Psi_1} + \frac{1}{\Omega_0} \right]^{k-l+1}} \left(1 - \sum_{v=1}^{M-v} \sum_{m=0}^{v-2} K_1 (-1)^m \right) - \frac{1}{\Omega_0} \sum_{v=1}^{M-v} \frac{1}{2} K_1 \frac{e^{-\left(\frac{2\Psi_t}{\Psi_1} + \frac{\sigma}{\Omega_0} \right)}}{\left(\frac{2\Psi_t}{\sigma\Psi_1} + \frac{1}{\Omega_0} \right)}$$

$$\text{where } K_1 = (-1)^{2v-1} \binom{M-v}{v}$$

Remark 1 1. p_{out_A} is the secrecy outage probability corresponding to GSC scheme at legitimate receiver B and MRC technique at eavesdropper ER with a primary user and single antenna Alice in Rayleigh fading environment.

2. P_{out_D} is the SOP corresponding to SC technique at B and MRC scheme at ER.

3. For $M_c = M$, then p_{out_A} is SOP corresponding to MRC scheme at both B and ER with single antenna Alice and a PU.

4. $P_{out_v}(\Psi_t)$ is the SOP as a function of threshold SNR corresponding to GSC scheme at legitimate receiver B and MRC technique at ER with a primary user. Similarly, $P_{out_1}(\Psi_t)$ is the SOP as a function of threshold SNR corresponding to SC technique at B and MRC scheme at ER.

The average number of active branches N_b with MS-GSC is expressed as

$$N_b = \sum_{n=1}^{M_c} n Pr [N_b = n] \sum_{n=1}^{M_c-1} p\gamma_n(\Psi_t) \quad (25)$$

where $nPr [N_b = n]$ is the PMF (probability mass function) of N_b and given by

$$Pr [N_b = n] = \begin{cases} 1 - P\gamma_1(\Psi_t) & n = 1 \\ P\gamma_{n-1}(\Psi_t) - P\gamma_n(\Psi_t) & 2 \leq n \leq M_c - 1 \\ P\gamma_{M_c-1}(\Psi_t) & n = M_c \end{cases} \quad (26)$$

3.1 Asymptotic secrecy outage probability

Here, asymptotic nature of SOP in high SNR regime $\psi_1 \rightarrow \infty$ is considered. By applying ([31], Eq. 1.211.1), the first-order expansion of $F_{\psi_1}^\infty$ is written as (27). Using (27) in (7), asymptotic SOP is calculated in (28). Here, we assume there are N_E antennas at eavesdropper, i.e., $M = N_E$.

$$F_{\psi_1|X}^\infty(x_1) = \begin{cases} \frac{1}{M_C^{M-M_C} M_C!} \left(\frac{x_1}{\psi_1}\right)^M, & 0 \leq x_1 \leq \psi_t \\ \vdots \\ \frac{\left(\frac{x_1}{\psi_1}\right)^M - P_{\psi_1}(\psi_t) + \frac{1}{M_C^{M-M_C} M_C!}}{\frac{(-1)^j M!}{(M-i-j)!(i-1)!(i-2)!(j)! \psi_1^j}} \\ \frac{\int_{\frac{(i-1)x_1}{i}}^{\frac{y}{i}} \int_{\psi_t-y}^{\frac{y}{i-1}} [y + (1-i)x_1]^{(i-2)} dzdy + \int_{\frac{(i-1)\psi_t}{i}}^{\psi_t} \int_{\psi_t-y}^{x_1-y} [y + (1-i)x_1]^{(i-2)} dzdy, \Psi_t \leq x_1 < \frac{M_C}{M_C-1} \Psi_t \\ \vdots \\ \frac{\left(\frac{x_1}{\psi_1}\right)^M - P_{\psi_1}(\psi_t) + \frac{1}{M_C^{M-M_C} M_C!}}{\frac{(-1)^j M!}{(M-i-j)!(i-1)!(i-2)!(j)! \psi_1^j}} \\ \frac{\int_{\frac{(i-1)x_1}{i}}^{\frac{y}{i}} \int_{\psi_t-y}^{\frac{y}{i-1}} [y + (1-i)x_1]^{(i-2)} dzdy + \int_{\frac{(i-1)\psi_t}{i}}^{\psi_t} \int_{\psi_t-y}^{x_1-y} [y + (1-i)x_1]^{(i-2)} dzdy, \frac{\nu+1}{\nu} \Psi_t \leq x_1 < \frac{\nu}{\nu-1} \Psi_t \\ \vdots \\ \left(\frac{x_1}{\psi_1}\right)^M, & 2\Psi_t < x_1 \end{cases} \quad (27)$$

$$P_{\psi_1}(\psi_t) = \sum_{i=0}^M \binom{M}{i} e^{-\frac{\psi_t}{\psi_1}}$$

$$P_{out}^{Asmptotic} = \begin{cases} P_A, & 0 \leq x_1 < \Psi_t; \\ \vdots \\ P_B - P_{out1}(\Psi_t) + P_A + \int_0^\infty \int_0^\infty \frac{1}{\Omega_0} e^{-\frac{y}{\Omega_0}} \frac{x^{N_E-1} e^{-\frac{x}{\psi_2}}}{\psi_2^{N_E} (N_E-1)!} \left\{ \sum_{i=2}^{M_C} \sum_{j=0}^{M-i} \frac{(-1)^j M!}{(M-i-j)!(i-1)!(i-2)!(j)! \psi_1^j} \int_{\frac{(i-1)x_1}{i}}^{\frac{y}{i}} \int_{\psi_t-y}^{\frac{y}{i-1}} [y + (1-i)x_1]^{(i-2)} dzdy + \int_{\frac{(i-1)\psi_t}{i}}^{\psi_t} \int_{\psi_t-y}^{x_1-y} [y + (1-i)x_1]^{(i-2)} dzdy \right\} dx dy, \Psi_t \leq x_1 < \frac{M_C}{M_C-1} \Psi_t \\ \vdots \\ P_B - P_{out1}(\Psi_t) + P_A + \int_0^\infty \int_0^\infty \frac{1}{\Omega_0} e^{-\frac{y}{\Omega_0}} \frac{x^{N_E-1} e^{-\frac{x}{\psi_2}}}{\psi_2^{N_E} (N_E-1)!} \left\{ \sum_{i=2}^l \sum_{j=0}^{M-i} \frac{(-1)^j M!}{(M-i-j)!(i-1)!(i-2)!(j)! \psi_1^j} \int_{\frac{(i-1)x_1}{i}}^{\frac{y}{i}} \int_{\psi_t-y}^{\frac{y}{i-1}} [y + (1-i)x_1]^{(i-2)} dzdy + \int_{\frac{(i-1)\psi_t}{i}}^{\psi_t} \int_{\psi_t-y}^{x_1-y} [y + (1-i)x_1]^{(i-2)} dzdy \right\} dx dy, \frac{\nu+1}{\nu} \Psi_t \leq x_1 < \frac{\nu}{\nu-1} \Psi_t \\ \vdots \\ P_B, & 2\Psi_t < x_1 \end{cases} \quad (28)$$

$$P_A = \frac{1}{M_C^{M-M_C} M_C!} P_B \quad (29)$$

where

$$P_B = \frac{\left(1 - e^{-\frac{\sigma}{\Omega_0}}\right)}{(\psi_2)^{N_E} (N_E-1)} \sum_{k=0}^M \binom{M}{k} \left(\frac{2^{R_s} - 1}{\psi_1}\right)^k \left(\frac{2^{R_s}}{\psi_1}\right)^{M-k} \int_0^\infty x^{N_E-1+M-k} e^{-\frac{x}{\psi_2}} dx + \sum_{k=0}^M \binom{M}{k} \left(\frac{2^{R_s} - 1}{\sigma \psi_1}\right)^k \left(\frac{2^{R_s}}{\sigma \psi_1}\right)^{M-k} \int_0^\infty x^{N_E-1+M-k} e^{-\frac{xy}{\sigma \psi_2}} dx \frac{1}{\Omega_0} \int_\sigma^\infty e^{-\frac{y}{\Omega_0}} dy \quad (30)$$

(28) can also be written as

$$P_{out}^{Asmptotic} = (G_A \psi_1)^{-G_D} + O\left(\psi_1^{-G_D}\right) \quad (31)$$

For $0 \leq x_1 \leq \psi_t$, the secrecy diversity order is

$$G_D = M, \quad (32)$$

and the secrecy array gain is

$$G_A = \left[\frac{1}{(M_C)! M_C^{M-M_C}} \frac{(q+N_E-1)!}{(N_E-1)!} \sum_{q=0}^M \binom{M}{q} (2^{R_s} - 1)^{M-q} (2^{R_s})^q \gamma_2^q \left(1 - e^{-\frac{\sigma}{\Omega_0}}\right) + \frac{1}{(M_C)! M_C^{M-M_C}} \frac{(q+N_E-1)!}{(N_E-1)!} \sum_{q=0}^M \binom{M}{q} (2^{R_s} - 1)^{M-q} (2^{R_s})^q (\sigma \gamma_2)^q e^{-\frac{\sigma}{\Omega_0}} \sum_{t=0}^{M-1} \frac{(M-q)}{t!} \left(\frac{\sigma}{\Omega_0}\right)^t \left(\frac{1}{\Omega_0}\right)^{M-q-t+1} \right]^{-\frac{1}{M}} \quad (33)$$

For $2\Psi_t < x_1$, the secrecy diversity order is $G_D = M$ and secrecy array gain is

$$G_{A1} = \left[\frac{(q+N_E-1)!}{(N_E-1)!} \sum_{q=0}^M \binom{M}{q} (2^{R_s} - 1)^{M-q} (2^{R_s})^q \psi_2^q \left(1 - e^{-\frac{\sigma}{\Omega_0}}\right) + \frac{(q+N_E-1)!}{(N_E-1)!} \sum_{q=0}^M \binom{M}{q} (2^{R_s} - 1)^{M-q} (2^{R_s})^q (\sigma \psi_2)^q e^{-\frac{\sigma}{\Omega_0}} \sum_{t=0}^{M-1} \frac{(M-q)}{t!} \frac{(\psi)^t}{\left(\frac{1}{\Omega_0}\right)^{M-q-t+1}} \right]^{-\frac{1}{M}} \quad (34)$$

According to (31), (32), (33), and (34), we have the following remarks to provide insight into the use of MS-GSC at secondary receiver.

Remark 2 As indicated in (28), asymptotic SOP approaches to zero as $\psi_1 \rightarrow \infty$. Furthermore, one also can observe that all asymptotic curves tightly approximate the exact curves in high ψ_1 regime.

Remark 3 The asymptotic result confirm that the secrecy diversity order is independent of N_E and ψ_2 , as indicated in (32). Note the secrecy outage probability increases with increasing N_E and ψ_2 . This confirms that the secrecy array gain in (33) and (34) is a decreasing function of N_E and ψ_2 .

Remark 4 The secrecy diversity order is independent of choice of M_c i.e. the number of combined antennas in B. It is dependent on the number of available receiver antennas at the B.

4 Numerical result

Numerical results highlight the effect of number of diversity branches, average SNR of ER, and σ on SOP. We assume $\Omega_0 = 1$ throughout this analysis. Exact and asymptotic curves are obtained from (15) and (28), respectively. These curves are also verified using Monte-Carlo simulations.

Figure 1 plots SOP of MS-GSC as given in (15) as a function of output threshold R_s for different numbers of diversity branches M_c . Here, we assume $\Psi_1 = 30$ dB, $\Psi_2 = 10$ dB, $\Psi_t = 7$ dB, $\sigma = 1$, and $M = 7$. We also plot SOP of 7/3-GSC for comparison. It is clear from the figure that SOP decreases as the number of diversity branches M_c increases, especially when the output threshold R_s is less than Ψ_t . When $R_s > \Psi_t$, the MS-GSC combiner try to increase the Ψ_{com} above Ψ_t and the SOP degrades. So, it is confirmed that the output threshold R_s should be less than or equal to the threshold SNR Ψ_t . For $M_c = 3$, SOP of MS-GSC is same as conventional GSC if $R_s > \Psi_t$.

Figure 2 plots SOP versus average SNR of the main channel Ψ_1 for different numbers of MS-GSC branches. As we see, SOP decreases as the number of MS-GSC branches M_c increases.

Figure 3 plots the exact and asymptotic SOP obtained from (15) and (28) versus the average SNR of main channel Ψ_1 for different numbers of MS-GSC branches with variation in average SNR of ER's channel. From the figure, it is clear that asymptotic curves correlate with the exact result at high SNR regime. According to (32), the secrecy diversity order is independent of N_E and only dependent on M . As Ψ_2 increases, secrecy performance degrades.

Figure 4 plots SOP for different values of σ . We see that SOP decreases with increasing σ . This is because of peak interference power constraint $\sigma = \frac{\Psi_p}{\Psi_0} = \frac{P_I}{P_{av}}$, which leads to increase in average transmit power of Alice given by (8).

Figure 5 plots active MRC branches versus threshold SNR Ψ_t . We can see that active MRC branches increases as threshold SNR ψ_t increases because it is very hard to increase the SNR of the combined Ψ_{com} above Ψ_t . MS-GSC requires less active MRC branches as compared to GSC that result in less processing power. The percentage

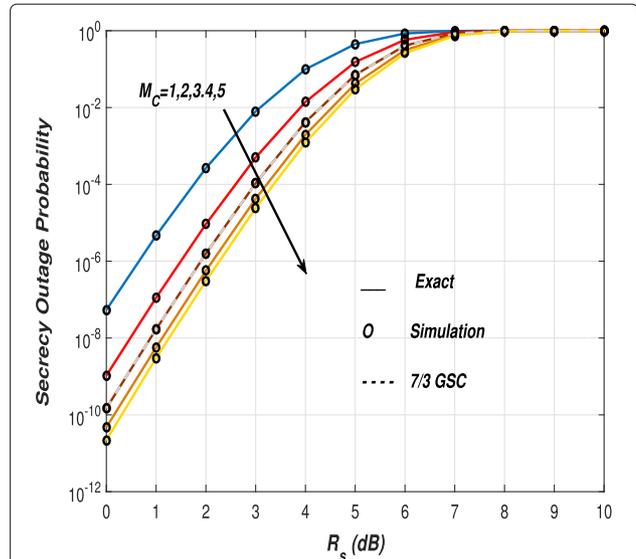


Fig. 1 Secrecy outage probability versus output threshold R_s . Secrecy outage probability versus R_s for different numbers of selected antennas M_c with $\psi_1 = 30$ dB, $\Psi_2 = 10$ dB, $\Psi_t = 7$ dB, $\sigma = 0.5$, and $M = 7$. Solid line represents the exact result and block dots represent the simulation result

of power saving for different number of antenna is shown in Table 1.

5 Conclusions

We proposed MS-GSC/MRC protocol for underlay cognitive radio network. We consider SOP as a main parameter to analyze secrecy of proposed system. We derived closed-form expression for exact and asymptotic SOP. From

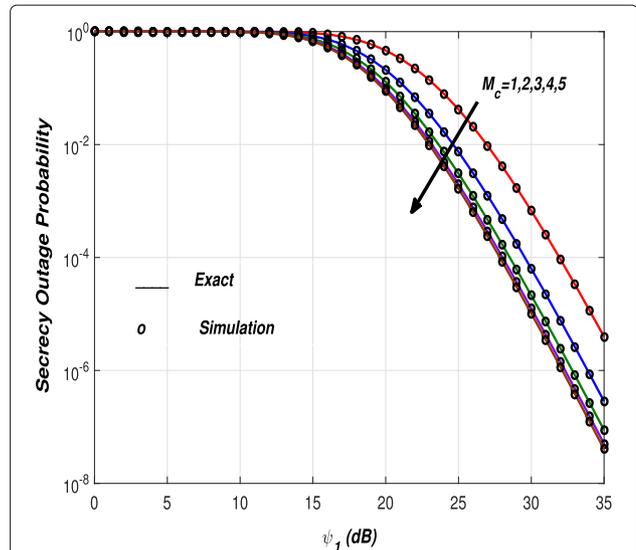
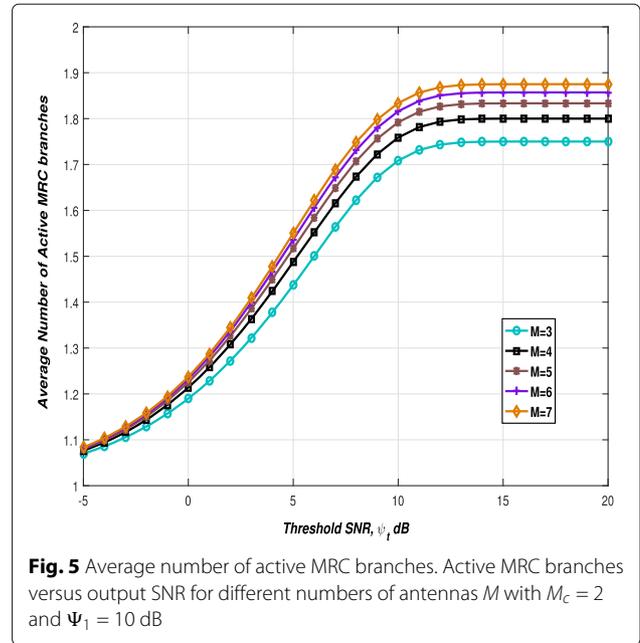
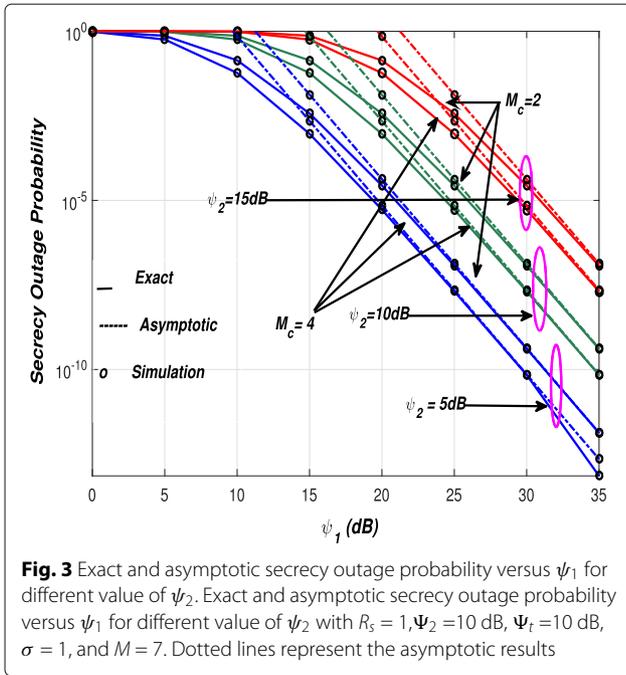


Fig. 2 Secrecy outage probability versus ψ_1 . Secrecy outage probability versus SNR of the main channel ψ_1 for different numbers of selected antennas M_c with $R_s = 1$, $\Psi_2 = 10$ dB, $\Psi_t = 10$ dB, $\sigma = 1$, and $M = 7$



numerical results, it has been concluded that MS-GSC saves more processing power in comparison to conventional GSC, and the number of selected antennas M_c and σ has positive impact on SOP.

Appendix

Based on (9), $Y \leq \frac{\Psi_p}{\Psi_0}, \Psi_M = \Psi_0 X_M, \Psi_E = \Psi_0 X_E$ and when $Y > \frac{\Psi_p}{\Psi_0}, \Psi_M = \frac{\Psi_p}{Y} X_M, \Psi_E = \frac{\Psi_p}{Y} X_E$. Hence, SOP for proposed system can be calculated as

$$P_{out} = \underbrace{\int_0^\sigma \int_0^\infty F_{\psi_1(X_1=x_1)}(x_1) f_{\psi_2(X=x)}(x) f_Y(y) dx dy}_{J_1} \underbrace{\int_\sigma^\infty \int_0^\infty F_{\psi_1(X_1=x_1)}(x_1) f_{\psi_2(X=x)}(x) f_Y(y) dx dy}_{J_2} \quad (35)$$

$$J_1 = \int_0^\sigma \int_0^\infty F_{\psi_1(X_1=x_1)}(x_1) f_{\psi_2(X=x)}(x) f_Y(y) dx dy \quad (36)$$

for $X_1 \leq \frac{\psi_p}{\psi_0}$, substituting Eqs. (10), (12), and (14) in Eq. (36), J_1 can be calculated as

$$J_1 = \begin{cases} P_{outA1} & 0 \leq x_1 < \Psi_t; \\ P_{outB1} & \Psi_t \leq x_1 < \frac{M_c}{M_c-1} \Psi_t \\ \vdots & \\ P_{outC1} & \frac{\nu+1}{\nu} \Psi_t \leq x_1 < \frac{\nu}{\nu-1} \Psi_t \\ \vdots & \\ P_{outD1} & 2\Psi_t < x_1 \end{cases} \quad (37)$$

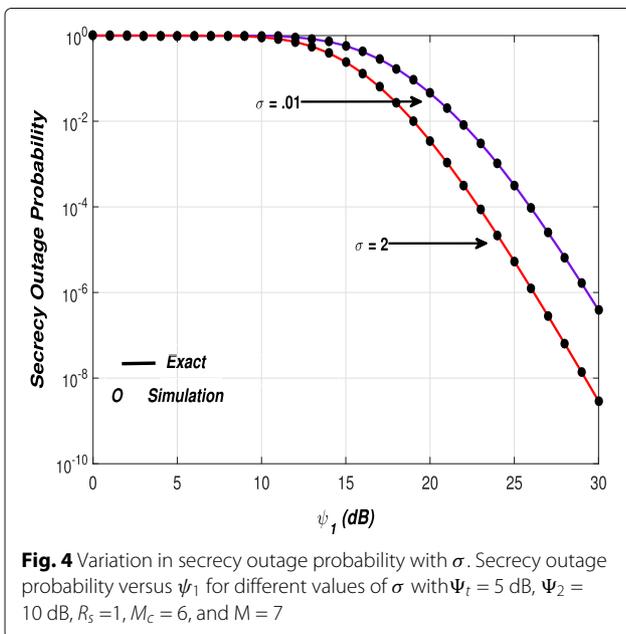


Table 1 Power saving by MS-GSC

Case	Number of antenna	M_c	Active MRC branches	Power saving (%)
1	3	2	1.750	12.5
2	4	2	1.800	10
3	5	2	1.833	8.35
4	6	2	1.857	7.15
5	7	2	1.875	6.25

$$P_{out_{A1}} = \frac{M!}{M_C!(M-M_C)!} \left\{ \left(1 - e^{-\frac{\sigma}{\Omega_0}}\right) \left(1 - \sum_{k=0}^{L_c-1} \sum_{n=0}^k \mu_1 e^{-\left(\frac{2^{R_s}-1}{\Psi_1}\right)}\right) + \sum_{v=1}^{M-M_c} C_2 C_3 - \sum_{v=1}^{M-M_c} \sum_{m=0}^{M_c-2} C_2 C_4 - \sum_{v=1}^{M-M_c} C_2 C_4 \beta_1 e^{-\left(\frac{(2^{R_s}-1)(1+\frac{v}{M_c})}{\Psi_1}\right)} + \sum_{v=1}^{M-M_c} \sum_{m=0}^{M_c-2} \sum_{k=0}^m \sum_{n=0}^k C_2 C_4 \mu_1 e^{-\left(\frac{2^{R_s}-1}{\Psi_1}\right)} \right\} \quad (38)$$

$$P_{out_{B1}} = P_{out_{D1}} - P_{out_{1A}}(\Psi_t) + P_{out_{M_{c1}}}(\Psi_t) + \sum_{w=2}^{M_c} P_{out_{M1}} \quad (39)$$

$$P_{out_M} = I_1 \quad (40)$$

$$P_{out_{D1}} = \left(1 - e^{-\frac{\sigma}{\Omega_0}}\right) \sum_{v=0}^M \binom{M}{v} (-1)^v \frac{e^{-\left(\frac{-v(2^{R_s}-1)}{\Psi_1}\right)}}{(\Psi_2)^M \left(\frac{v2^{R_s}}{\Psi_1} + \frac{1}{\Psi_2}\right)^M} \quad (41)$$

$$P_{out_{1A}}(\Psi_t) = \sum_{v=0}^M \binom{M}{v} (-1)^v e^{-\frac{v\Psi_t}{\Psi_1}} \left(1 - e^{-\frac{\sigma}{\Omega_0}}\right) \quad (42)$$

$$P_{out_{M_{c1}}}(\Psi_t) = \left(1 - e^{-\frac{\sigma}{\Omega_0}}\right) \frac{M!}{(M-M_c)!M_c!} \left\{ 1 - e^{-\frac{\Psi_t}{\Psi_1}} \sum_{k=0}^{M_c-1} \frac{1}{k!} \left(\frac{\Psi_t}{\Psi_1}\right)^k + \sum_{v=1}^{M-M_c} C_2 \left[C_3 \left(1 - e^{-\left(1+\frac{v}{M_c}\right)\frac{\Psi_t}{\Psi_1}}\right) - \sum_{m=0}^{M_c-2} C_4 \left(1 - e^{-\frac{\Psi_t}{\Psi_1}} \sum_{k=0}^m \frac{1}{k!} \left(\frac{\Psi_t}{\Psi_1}\right)^k\right) \right] \right\} \quad (43)$$

$$P_{out_{C1}} = P_{out_D} - P_{out_{1A}}(\Psi_t) + P_{out_{v1}}(\Psi_t) + \sum_{w=2}^v P_{out_{M1}} \quad (44)$$

$$P_{out_{v1}}(\Psi_t) = \left(1 - e^{-\frac{\sigma}{\Omega_0}}\right) \frac{M!}{(M-v)!v!} \left\{ 1 - e^{-\frac{\Psi_t}{\Psi_1}} \sum_{k=0}^{v-1} \frac{1}{k!} \left(\frac{\Psi_t}{\Psi_1}\right)^k + \sum_{v=1}^{M-v} (-1)^{2v-1} \frac{(M-v)!}{(M-v-v)!v!} \times \left[\frac{1}{2} \left[1 - e^{-\frac{2\Psi_t}{\Psi_1}} \right] - \sum_{m=0}^{v-2} (-1)^m \left(1 - e^{-\frac{\Psi_t}{\Psi_1}} \sum_{k=0}^m \frac{1}{k!} \left(\frac{\Psi_t}{\Psi_1}\right)^k\right) \right] \right\} \quad (45)$$

$$I_1 = \left(1 - e^{-\frac{\sigma}{\Omega_0}}\right) \left(r_1 \left[r_5 - \frac{e^{-\frac{2^{R_s}-1}{\Psi_1}} \Psi_2^M \sum_{a=0}^{m-v} (2^{R_s}-1)^{m-v-a} (2^{R_s})^a (a+M-1)!}{(M-1)! \left(\frac{2^{R_s}}{\Psi_1} + \frac{1}{\Psi_2}\right)^{a+M}} + \frac{e^{-\frac{2^{R_s}-1}{\Psi_1}}}{(M-1)! \Psi_2^M} \sum_{a=0}^{m-l} (2^{R_s}-1)^{m-v-a} (2^{R_s})^a \sum_{b=0}^{w-1-k+v} \binom{w-1-k+v}{b} \left(\frac{(w-1)(2^{R_s}-1)}{w\Psi_t}\right)^{(w-1-k+v-b)} \left(\frac{(w-1)(2^{R_s})}{w\Psi_t}\right)^b \frac{(a+b+M-1)!}{\left(\frac{2^{R_s}}{\Psi_1} + \frac{1}{\Psi_2}\right)^{a+b+M}} \right] + r_2 \left[r_3 - \sum_{a=0}^{m-v} \sum_{n=0}^{w-2-k+v} \binom{m-v}{a} (2^{R_s}-1)^{m-v-a} (2^{R_s})^a e^{-\left(\frac{(w+j)(2^{R_s}-1)}{w\Psi_1}\right)} \binom{n}{a} \frac{1}{n!} \left(\frac{-(w-1)j(2^{R_s}-1)}{w\Psi_1}\right)^{n-b} \left(\frac{-(w-1)j(2^{R_s})}{w\Psi_1}\right)^n \frac{(a+b+M-1)!}{(M-1)! \Psi_2^M} \frac{1}{\left(\left(\frac{(w+j)(2^{R_s})}{\Psi_1}\right) + \frac{1}{\Psi_2}\right)^{a+b+M}} + \sum_{a=0}^{m-v} \sum_{n=0}^{w-2-k+v} \binom{m-v}{a} (2^{R_s}-1)^{m-v-a} (2^{R_s})^a e^{-\left(\frac{-(j+1)(2^{R_s}-1)}{\Psi_1}\right)} \frac{(a+M-1)!}{(M-1)! \Psi_2^M} \frac{1}{\left(\left(\frac{(w+j)(2^{R_s})}{\Psi_1}\right) + \frac{1}{\Psi_2}\right)^{a+b+M}} \right] + r_4 \left[r_{10} - \sum_{n=0}^{w-2-k+m} \sum_{b=0}^n e^{-\left(\frac{-(w+j)(2^{R_s}-1)}{w\Psi_1}\right)} e^{-\left(\frac{-(w+j)(2^{R_s})}{w\Psi_1}\right)} \frac{1}{n!} \left(\frac{(w+j)}{w\Psi_1}\right)^n \binom{n}{b} (2^{R_s}-1)^{n-b} \frac{(2^{R_s})^b (a+M-1)!}{(M-1)! \Psi_2^M \left(\frac{(w+j)2^{R_s}}{w\Psi_1} + \frac{1}{\Psi_2}\right)^{b+M}} \right] \right) \quad (46)$$

$$r_1 = \frac{M!}{(M-w)!(w-1)!} \sum_{k=0}^{w-2} \frac{(1-w)^k}{(w-2-k)!} \sum_{m=0}^k \frac{1}{\Psi_1^{w-1-k+m}} \sum_{v=0}^m \frac{(-1)^v}{(m-v)!} \frac{\Psi_t^{w-1-k+v}}{w-1-k+v}, \quad r_5 = e^{-\frac{\Psi_t}{\Psi_1}} \Psi_t^{m-v} \left(1 - \left(\frac{w-1}{w}\right)^{w-1-k+v}\right)$$

$$r_2 = \sum_{j=1}^{M-w} \frac{(-1)^j M!}{(L-w-j)!(w-1)!j! \Psi_1^w} \sum_{k=0}^{w-2} \frac{(1-w)^k}{(w-2-k)!} \sum_{m=0}^k \left(\frac{j+1}{\gamma_1}\right)^{m-k-1} \sum_{v=0}^m \frac{(-1)^l (w-2-k+l)!}{(m-v)!v!} \left(\frac{-\Psi_t}{j}\right)^{w-1-k+v}$$

$$r_3 = \Psi_t^{m-v} \left(e^{-\frac{(w+j)\Psi_t}{w\Psi_1}} \sum_{n=0}^{i-2-k+l} \frac{1}{n!} \left(\frac{-(w-1)j\Psi_t}{w\Psi_1}\right)^n - e^{-\frac{\Psi_t}{\Psi_1}} \sum_{n=0}^{w-2-k+l} \frac{1}{n!} \left(\frac{-j\Psi_t}{\Psi_1}\right)^n \right)$$

$$r_4 = \sum_{j=0}^{M-w} \frac{(-1)^j M!}{(M-w-j)!(w-1)!j! \Psi_1^w} \sum_{k=0}^{w-2} \frac{(1-w)^k}{(w-2-k)!} \sum_{m=0}^k \frac{(w-2-k+m)!}{m!(w-1)^m} \left(\frac{j+1}{\Psi_1}\right)^{m-k-1} \left(\frac{(w-1)\Psi_1}{w+j}\right)^{w-1-k+m}$$

$$r_{10} = e^{-\frac{(w+j)\Psi_t}{w\Psi_1}} \sum_{n=0}^{w-2-k+m} \frac{1}{n!} \left(\frac{(w+j)\Psi_t}{w\Psi_1}\right)^n, \quad r_{25} = \frac{1}{n!} \left(\frac{-j\Psi_t}{\Psi_1}\right)^n$$

I_1 given in (20) can be calculated using ([22], Eq. (30)).

$$J_2 = \int_{\frac{\psi_p}{\gamma_0}}^{\infty} \int_0^{\infty} F_{\gamma_1(X_1=x_1)}(x_1) f_{\gamma_2(X=x)}(x) f_Y(y) dx dy \quad (47)$$

for $X_1 > \frac{\gamma_p}{\gamma_0}$

By Eqs. (10), (12), and (14) in Eq. (36) and using

$$\int_{\rho}^{\infty} x^m e^{-\mu x} dx = e^{-\rho\mu} \sum_{p=0}^m \frac{m!}{p!} \frac{\rho^k}{\mu^{m-p+1}}, J_2 \text{ can be given as}$$

$$J_2 = \begin{cases} P_{\text{out}_{A2}} & 0 \leq x_1 < \Psi_t; \\ P_{\text{out}_{B2}} & \Psi_t \leq x_1 < \frac{M_c}{M_c-1} \Psi_t \\ \vdots & \\ P_{\text{out}_{C2}} & \frac{\nu+1}{\nu} \Psi_t \leq x_1 < \frac{\nu}{\nu-1} \Psi_t \\ \vdots & \\ P_{\text{out}_{D2}} & 2\Psi_t < x_1 \end{cases} \quad (48)$$

$$P_{\text{out}_{A2}} = \frac{M!}{M_c! (M - M_c)!} \left\{ e^{-\frac{\sigma}{\Omega_0}} - \sum_{k=0}^{M_c-1} \sum_{n=0}^k \sum_{p=0}^{k-n} Q_1 Q_2 e^{-\sigma \left(\frac{2^{Rs}-1}{\sigma\psi_1} + \frac{1}{\Omega_0} \right)} + \sum_{\nu=1}^{M-M_c} C_2 C_3 e^{-\left(\frac{\sigma}{\Omega_0} \right)} - \sum_{\nu=1}^{M-M_c} C_2 C_3 \beta_2 e^{-\sigma \left(\left(1 + \frac{M}{M_c} \right) \left(\frac{2^{Rs}-1}{\sigma\psi_1} \right) + \frac{1}{\Omega_0} \right)} - \sum_{\nu=1}^{M-M_c} \sum_{m=0}^{M_c-2} C_2 C_4 e^{-\left(\frac{\sigma}{\Omega_0} \right)} + \sum_{\nu=1}^{M-M_c} \sum_{m=0}^{M_c-2} \sum_{k=0}^m \sum_{n=0}^k \sum_{p=0}^{k-n} C_2 C_4 Q_1 Q_2 e^{-\sigma \left(\frac{2^{Rs}-1}{\sigma\psi_1} + \frac{1}{\Omega_0} \right)} \right\}$$

$$P_{\text{out}_{B2}} = P_{\text{out}_{D2}} - P_{\text{out}_{2A}}(\Psi_t) + P_{\text{out}_{M_c2}}(\Psi_t) + \sum_{w=2}^{M_c} P_{\text{out}_M} \quad (49)$$

$$P_{\text{out}_{D2}} = \sum_{\nu=0}^M \binom{M}{\nu} (-1)^\nu \times \left(\frac{1}{(\sigma\psi_2)^M \left(\frac{\nu 2^{Rs}}{\sigma\psi_1} + \frac{1}{\sigma\psi_2} \right)^M} \frac{e^{-\left(\frac{\nu(2^{Rs}-1)}{\psi_1} + \frac{\sigma}{\Omega_0} \right)}}{\left(\frac{\nu(2^{Rs}-1)}{\sigma\psi_1} + \frac{1}{\Omega_0} \right)} \right) \quad (50)$$

$$P_{\text{out}_{2A}}(\Psi_t) = \sum_{\nu=0}^M \binom{M}{\nu} (-1)^\nu \frac{e^{-\sigma \left(\frac{\nu\psi_t}{\psi_1} + \frac{\sigma}{\Omega_0} \right)}}{\left(\frac{\nu\psi_t}{\sigma\psi_1} + \frac{1}{\Omega_0} \right)} \quad (51)$$

$$P_{\text{out}_{M_c2}}(\Psi_t) = \frac{M!}{(M - M_c)! M_c!} \left[e^{-\frac{\sigma}{\Omega_0}} \left(1 + \sum_{\nu=1}^{M-M_c} C_2 C_3 - \sum_{\nu=1}^{M-M_c} \sum_{m=0}^{M_c-2} C_2 C_4 \right) - \frac{1}{\Omega_0} \sum_{k=0}^{M_c-1} \sum_{l=0}^k \frac{1}{k!} \left(\frac{\psi_t}{\sigma\psi_1} \right)^k e^{-\left(\frac{\psi_t}{\psi_1} + \frac{\sigma}{\Omega_0} \right)} \frac{k!}{l!} \frac{(\sigma)^l}{\left(\frac{\psi_t}{\sigma\psi_1} + \frac{1}{\Omega_0} \right)^{k-l+1}} \left(1 - \sum_{\nu=1}^{M-M_c} \sum_{m=0}^{M_c-2} C_2 C_4 \right) - \frac{1}{\Omega_0} \sum_{\nu=1}^{M-M_c} C_2 C_3 \frac{e^{-\left(\frac{\psi_t}{C_3\sigma\psi_1} + \frac{\sigma}{\Omega_0} \right)}}{\frac{\psi_t}{C_3\sigma\psi_1} + \frac{\sigma}{\Omega_0}} \right] \quad (52)$$

$$P_{\text{out}_{C2}} = P_{\text{out}_{D2}} - P_{\text{out}_1}(\Psi_t) + P_{\text{out}_{\nu 2}}(\Psi_t) + \sum_{w=2}^{\nu} P_{\text{out}_{M1}} \quad (53)$$

$$P_{\text{out}_{\nu 2}}(\Psi_t) = \frac{M!}{(M - \nu)! \nu!} \left[e^{-\frac{\sigma}{\Omega_0}} \left(1 + \sum_{\nu=1}^{M-\nu} \frac{1}{2} K_1 - \sum_{\nu=1}^{M-\nu} \sum_{m=0}^{\nu-2} K_1 (-1)^m \right) - \frac{1}{\Omega_0} \sum_{k=0}^{\nu-1} \sum_{l=0}^k \frac{1}{k!} \left(\frac{\psi_t}{\sigma\psi_1} \right)^k e^{-\left(\frac{\psi_t}{\psi_1} + \frac{\sigma}{\Omega_0} \right)} \frac{k!}{l!} \frac{(\sigma)^l}{\left[\frac{\psi_t}{\psi_1\sigma} + \frac{1}{\Omega_0} \right]^{k-l+1}} \left(1 - \sum_{\nu=1}^{M-\nu} \sum_{m=0}^{\nu-2} K_1 (-1)^m \right) - \frac{1}{\Omega_0} \sum_{\nu=1}^{M-\nu} \frac{1}{2} K_1 \frac{e^{-\left(\frac{2\psi_t}{\psi_1} + \frac{\sigma}{\Omega_0} \right)}}{\left(\frac{2\psi_t}{\sigma\psi_1} + \frac{1}{\Omega_0} \right)} \right] \quad (54)$$

$$P_{\text{out}_{M2}} = I_2 \quad (55)$$

$$\begin{aligned}
 I_2 = & \left(\left[r'_1 r'_5 \sum_{c=0}^{w-1-k+m} \frac{(w-1-k+m)!}{c!} \frac{(\sigma)^c e^{\left(-\frac{\sigma \Psi_t}{\sigma \Psi_1 + \frac{1}{\Omega_0}}\right)}}{\left(\frac{\Psi_t}{\sigma \Psi_1} + \frac{1}{\Omega_0}\right)^{w-k+m-c}} - \frac{r'_1}{(M-1)! (\sigma \Psi_1)^M} \sum_{a=0}^{m-v} (2^{Rs}-1)^{m-v-a} (2^{Rs})^a \frac{(a+M-1)!}{\left(\frac{2^{Rs}}{\sigma \Psi_1} + \frac{1}{\sigma \Psi_2}\right)^{a+M}} \binom{m-v}{a} \right. \right. \\
 & \left. \left. e^{-\sigma \left(\frac{2^{Rs}-1}{\sigma \Psi_1} + \frac{1}{\Omega_0}\right)} \sum_{d=0}^{w-1-k+m+a} \frac{(w-1-k+m+a)!}{d!} \frac{(\sigma)^d}{\left(\frac{2^{Rs}-1}{\sigma \Psi_1} + \frac{1}{\Omega_0}\right)^{w-k+m+a}} + \frac{r'_1}{(M-1)! (\sigma \Psi_1)^M} \sum_{a=0}^{m-v} \binom{m-v}{a} \right. \right. \\
 & \left. \left. (2^{Rs}-1)^{m-v-a} (2^{Rs})^a \sum_{b=0}^{w-1-k+l} \binom{w-1-k+l}{a} \left(\frac{(w-1)(2^{Rs}-1)}{s \Psi_t}\right)^{(w-1-k+v-b)} \left(\frac{(v-1)(2^{Rs})}{w \gamma_t}\right)^b \frac{(a+b+M-1)!}{\left(\frac{2^{Rs}}{\sigma \Psi_1} + \frac{1}{\sigma \Psi_2}\right)^{a+b+M}} \right. \right. \\
 & \left. \left. e^{-\sigma \left(\frac{2^{Rs}-1}{\sigma \Psi_1} + \frac{1}{\Omega_0}\right)} \sum_{e=0}^{w-1-k+m-a-b} \frac{(w-1-k+m-a-b)!}{e!} \frac{(\sigma)^e}{\left(\frac{2^{Rs}-1}{\sigma \Psi_1} + \frac{1}{\Omega_0}\right)^{w-k+m-a-b-e}} \right] \right. \\
 & + \left[r'_2 \Psi_t^{m-v} \sum_{n=0}^{w-2-k+v} \frac{1}{n!} \left(\frac{-(w-1)j \Psi_t}{\sigma \Psi_1 \Psi_t}\right)^n e^{-\sigma \left(\frac{(w+j) \Psi_t}{w \sigma \Psi_1} + \frac{1}{\Omega_0}\right)} \sum_{f=0}^{n+m-v} \frac{(n+m-v)!}{f!} \right. \\
 & \frac{(\sigma)^f}{\left(\frac{(w+j) \Psi_t}{w \sigma \Psi_1} + \frac{1}{\Omega_0}\right)} - r'_2 \Psi_t^{m-v} \sum_{n=0}^{w-2-k+l} \frac{1}{n!} \left(\frac{-j \Psi_t}{\sigma \Psi_1}\right)^n \sum_{g=0}^{n+m-v} \frac{(n+m-v)!}{g!} \frac{(\sigma)^g}{\left(\frac{\Psi_t}{\sigma \Psi_1} + \left(\frac{1}{\Omega_0}\right)\right)^{n+m-v-g+1}} \\
 & - r'_2 \sum_{a=0}^{m-v} \sum_{n=0}^{w-2-k+v} \sum_{b=0}^n \binom{m-v}{a} (2^{Rs}-1)^{m-v-a} (2^{Rs})^a \binom{n}{b} \frac{1}{n!} \left(\frac{-j(w-1)(2^{Rs}-1)}{w \sigma \Psi_1}\right)^{n-b} \left(\frac{-j(w-1)(2^{Rs})}{w \sigma \Psi_1}\right)^b \\
 & \frac{(a+b+M-1)!}{(M-1)! (\sigma \Psi_2)^M} \frac{1}{\left(\frac{(w+j) 2^{Rs}}{\sigma \Psi_1} + \frac{1}{\sigma \Psi_2}\right)^{a+b+M}} e^{-\sigma \left(\frac{(w+j)(2^{Rs}-1)}{w \sigma \Psi_1} + \frac{1}{\Omega_0}\right)} \sum_{h=0}^{n-a-b} \frac{(n-a-b)!}{h!} \frac{(\sigma)^h}{\left(\frac{(w+j)(2^{Rs}-1)}{w \sigma \Psi_1} + \frac{1}{\Omega_0}\right)^{n-a-b+h+1}} + \sum_{a=0}^{m-v} \\
 & \sum_{n=0}^{w-2-k+v} r'_2 \binom{m-v}{a} (2^{Rs}-1)^{m-v-a} (2^{Rs})^a r'_{25} \frac{(a+M-1)!}{(M-1)! (\Psi_2 \sigma)^M} \frac{1}{\left(\frac{(j+1)(2^{Rs})}{\sigma \Psi_1} + \frac{1}{\sigma \Psi_2}\right)^{a+M}} e^{-\sigma \left(\frac{(j+1)(2^{Rs}-1)}{\sigma \Psi_1} + \frac{j \Psi_t}{\sigma \Psi_1} + \frac{1}{\Omega_0}\right)} \sum_{q=0}^{m-v+n-a} \frac{(m-v+n-a)!}{q!} \\
 & \left. \frac{(\sigma)^p}{\left(\frac{(j+1)(2^{Rs}-1)}{\sigma \Psi_1} + \frac{j \Psi_t}{\sigma \Psi_1} + \frac{1}{\Omega_0}\right)^{m-v+n-a-q+1}} \right] + \left[\frac{r'_4}{(M-1)! (\sigma \Psi_2)^M} \sum_{n=0}^{w-2-k+m} \frac{1}{n!} \left(\frac{(w+j) \Psi_t}{w \sigma \Psi_1}\right)^n e^{-\sigma \left(\frac{(w+j) \Psi_t}{w \sigma \Psi_1} + \frac{1}{\Omega_0}\right)} \right. \\
 & \left. \sum_{r=0}^n \frac{n!}{r!} \frac{(\sigma)^r}{\left(\frac{(w+j) \Psi_t}{w \sigma \Psi_1} + \frac{1}{\Omega_0}\right)^{n-r+1}} - \frac{r'_4}{(M-1)! (\sigma \Psi_2)^M} \sum_{n=0}^{w-2-k+m} \sum_{b=0}^n \frac{1}{n!} \left(\frac{(w+j)}{w \sigma \Psi_1}\right)^n \binom{n}{b} \right. \\
 & \left. (2^{Rs}-1)^{n-b} (2^{Rs})^b \frac{(b+M-1)!}{\left(\frac{(w+j) 2^{Rs}}{w \sigma \Psi_1} + \frac{1}{\sigma \Psi_2}\right)^{b+M}} e^{-\sigma \left(\frac{(w+j)(2^{Rs}-1)}{w \sigma \Psi_1} + \frac{(w+j) 2^{Rs}}{w \sigma \Psi_1} + \frac{1}{\Omega_0}\right)} \sum_{s=0}^{n-b-M} \frac{(n-b-M)!}{s!} \frac{(\sigma)^s}{\left(\frac{(w+j)(2^{Rs}-1)}{w \sigma \Psi_1} + \frac{(w+j) 2^{Rs}}{w \sigma \Psi_1} + \frac{1}{\Omega_0}\right)^{n-b-v-s+1}} \right] \Bigg) \\
 \\
 r'_1 = & \frac{M!}{(M-w)! (w-1)!} \sum_{k=0}^{w-2} \frac{(1-w)^k}{(w-2-k)!} \sum_{m=0}^k \frac{1}{(\Psi_1 \sigma)^{w-1-k+m}} \sum_{v=0}^m \frac{(-1)^v}{(m-v)! v!} \frac{\Psi_t^{w-1-k+v}}{(w-1-k+l)}, \quad r'_5 = \Psi_t^{m-v} \left(1 - \left(\frac{v-1}{w}\right)^{w-1-k+v}\right) \\
 r'_2 = & \sum_{j=1}^{M-w} \frac{(-1)^j M!}{(M-w-j)! (w-1)! j! (\sigma \Psi_1)^w} \sum_{k=0}^{w-2} \frac{(1-w)^k}{(w-2-k)!} \sum_{m=0}^k \frac{(j+1)^{m-k-1}}{(\sigma \Psi_1)^m} \sum_{v=0}^m \frac{(-1)^l (w-2-k+v)!}{(m-v)! v!} \left(\frac{-\Psi_t}{j}\right)^{w-1-k+v} \\
 r'_4 = & \sum_{j=0}^{M-w} \frac{(-1)^j M!}{(M-w-j)! (w-1)! j! (\sigma \Psi_1)^w} \sum_{k=0}^{w-2} \frac{(1-w)^k}{(w-2-k)!} \sum_{m=0}^k \frac{(w-2-k+m)!}{m! (w-1)^m} \left(\frac{j+1}{\sigma \Psi_1}\right)^{m-k-1} \left(\frac{(w-1) \sigma \Psi_1}{w+j}\right)^{w-1-k+m}, \quad r'_{25} = \frac{1}{n!} \left(\frac{-j \Psi_t}{\sigma \Psi_1}\right)^n
 \end{aligned}$$

(56)

Abbreviations

CRN: Cognitive radio network; CSI: Channel state information; ER: Eavesdropper; GSC: Generalized-selection combining; MRC: Maximal ratio combining; MS-GSC: Minimum selection GSC; PU: Primary user; SOP: Secrecy outage probability; SR: Secondary receiver; SU: Secondary user; TAS: Transmit antenna selection

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Authors' contributions

Both authors examine the secrecy performance of an underlay CRN with minimum selection-generalized selection combining (MS-GSC) over Rayleigh fading environment and derived closed-form expression for exact and asymptotic secrecy outage probability. Both authors read and approved the final manuscript.

Competing interests

The authors declare that they have no competing interests.

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References

- N Yang, L Yeoh, M ElKashlan, Transmit antenna selection for security enhancement in MIMO wiretap channels. *IEEE Trans. Commun.* **61**(1), 144–154 (2013)
- M ElKashlan, L Wang, TQ Duong, et al, On the security of cognitive radio networks. *IEEE Trans. Veh. Technol.* **64**(8), 3790–3795 (2015)
- M Gastpar, On capacity under receive and spatial spectrum-sharing constraints. *IEEE Trans. Inf. Theory.* **53**(2), 471–487 (2007)
- A Goldsmith, SA Jafar, I Maric, Breaking spectrum gridlock with cognitive radios: an information theoretic perspective. *Proc. IEEE.* **64**, 3790–3795 (2009)
- A Mukherjee, AL Swindlehurst, Robust beamforming for security in MIMO wiretap channels with imperfect CSI. *IEEE Trans. Signal Process.* **59**, 351–361 (2011)
- J Huang, AL Swindlehurst, Cooperative jamming for secure communications in MIMO relay networks. *IEEE Trans. Signal Process.* **51**, 4871–4884 (2011)
- HV Poor, Information and inference in the wireless physical layer. *IEEE Wirel. Commun.* **19**(1) (2012). February 2012
- HE Fangming, M Hong, W Wang, Maximal ratio diversity combining enhanced security. *IEEE Commun. Lett.* **15**, 509–511 (2011)
- H Alves, RD Souza, M Debbah, Performance of transmit antenna selection physical layer security schemes. *IEEE Signal Process. Lett.* **19**, 372–375 (2012)
- N Yang, P Yeoh, M ElKashlan, Transmit antenna selection for security enhancement in MIMO wiretap channels. *IEEE Trans. Commun.* **61**, 144–154 (2013)
- J Huang, A Mukherjee, AL Swindlehurst, Secure communication via an untrusted non-regenerative relay in fading channels. *IEEE Trans. Signal Process.* **61**, 2536–2550 (2013)
- H Zhao, D Wang, C Tang, et al, Physical layer security of underlay cognitive radio using maximal ratio combining. *Front. Inf. Technol. Electron. Eng.* **17**(9), 929–937 (2016)
- H Liu, H Zhao, C Tang, et al, Physical-layer secrecy outage of spectrum sharing CR systems over fading channels. *Sci. China Inf. Sci.* **59**(10), 102–302 (2016)
- H Zhao, Y Tan, et al, Secrecy outage on transmit antenna selection/maximal ratio combining in MIMO cognitive radio networks. *IEEE Trans. Veh. Technol.* **65**(12), 6–10242 (2016)
- H Lei, M Xu, I Ansari, et al, On secure underlay MIMO cognitive radio networks with energy harvesting and transmit antenna selection. *IEEE Trans. Green Commun. Netw.* **1**(2), 192–203 (2017)
- H Lei, H Zhang, IS Ansari, et al, On secrecy outage of relay selection in underlay cognitive radio networks over Nakagami- m fading channels. *IEEE Trans. Cogn. Commun. Netw.* **3**(4), 614–627 (2017)
- G Pan, J Ye, et al, Secure hybrid VLC-RF systems with light energy harvesting. *IEEE Trans. Commun.* **65**(10), 4348–4359 (2017)
- C Tang, G Pan, et al, Secrecy outage analysis of underlay cognitive radio unit over Nakagami- m fading channels. *IEEE Wirel. Commun. Lett.* **3**(6), 609–612 (2014)
- C Zhang, G Wang, et al, Secrecy outage analysis on underlay cognitive radio system with full-duplex secondary user. *IEEE Wirel. Commun. Lett. IEEE Access.* **3**(6), 25696–25705 (2017)
- MK Simon, MS Alouini, Performance analysis of generalized selection combining with threshold test per branch (T-GSC). *IEEE Trans. Veh. Technol.* **51**(5), 1018–1029 (2002)
- A Annamalai, G Deora, C Tellambura, in *Wireless Communications and Networking, 2003. WCNC 2003. 2003 IEEE*. Unified analysis of generalized selection diversity with normalized threshold test per branch, vol. 2 (IEEE, 2003), pp. 752–756
- X Zhang, NC Beaulieu, in *Global Telecommunications Conference, 2004. GLOBECOM'04. IEEE*. Threshold-based hybrid selection/maximal-ratio combining over generalized fading channels, vol. 1 (IEEE, 2004), pp. 462–468
- H Lei, H Zhang, et al, Secrecy outage performance for SIMO underlay cognitive radio systems with generalized selection combining over Nakagami- m channels. *IEEE Trans. Veh. Technol.* **65**(12), 10126–10132 (2016)
- N Yang, PL Yeoh, et al, MIMO wiretap channels: secure transmission using transmit antenna selection and receive generalized selection combining. *IEEE Commun. Lett.* **17**(9), 1754–1757 (2013)
- L Wang, M ElKashlan, et al, Secure transmission with antenna selection in MIMO Nakagami- m fading channels. *IEEE Trans. Wirel. Commun.* **13**(11), 6054–6067 (2014)
- Y Deng, L Wang, M ElKashlan, et al, Generalized selection combining for cognitive relay networks over Nakagami- m fading. *IEEE Trans. Signal Process.* **63**(8), 1993–2006 (2015)
- P Gupta, N Bansal, RK Mallik, in *Communications, 2004 IEEE International Conference on*. Analysis of minimum selection GSC in Rayleigh fading, vol. 6 (IEEE, 2004), pp. 3364–3368
- HC Yang, New results on ordered statistics and analysis of minimum-selection generalized selection combining (GSC). *IEEE Trans. Wirel. Commun.* **5**(7), 1876–1885 (2006)
- J Lee, JG Wang, et al, Outage probability of cognitive relay networks with interference constraints. *IEEE Trans. Wirel. Commun.* **10**(2), 390–395 (2011)
- R Etkin, A Parekh, D Tse, Spectrum sharing for unlicensed bands. *IEEE J. Sel. Areas Commun.* **25**(2), 517–528 (2007)
- IS Gradshteyn, IM Ryzhik, *Table of integrals*, 7th ed. (Academic, San Diego, 2007)

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